

A PROJECT REPORT (THESIS)
ON
FPGA IMPLEMENTATION OF UTM
AND PROTOCOL LAYER FOR USB 2.0

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CERTIFICATE

This is to certify that the thesis entitled, “**FPGA IMPLEMENTATION OF UTMI AND PROTOCOL LAYER OF USB 2.0**” submitted by **SWARUP KUMAR MAHAPATRA** (109EC0222) and **SADHNA MUANKHIA** (109EC0224) in partial fulfillment of the requirements for the award of Bachelor of Technology degree in **Electronics and Communication Engineering** during the session 2012-13 at National Institute of Technology, Rourkela (Deemed University) and is a bonafied work done by them under my supervision and guidance.

To the best of my knowledge, the matter embodied in the thesis has not been submitted to any other university/institute for the award of any Degree/Diploma.

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ABSTRACT

With the advancement in hardware technology, FPGA coming into existence are able to accommodate faster and more powerful electronic devices on a single device.

IP Cores have been developed to enable design reuse, increasing functionality, performance and flexibility.

USB is a serial bus which can realize the Plug and Play feature for easy connection of peripherals to PCs. It is a point to point interface in which data rate of over 480Mbit/s can be transferred as per new USB 2.0 Specification. It provides bi-directional, low- cost and high speed serial interface for data transfer. Multiple devices can be attached through a hub to the host. The USB Communication implemented complies with USB 2.0 Specifications essential for basic data transfer and can operate at USB Full speed (12 Mbit/s) and High speed (480Mbit/s).

This project deals with implementation of a USB Core specifically UTMI and protocol layer module on FPGA. The design is done by writing code in verilog HDL which is then verified and synthesized using Xilinx XST. The result is verified using testbench waveform to study the transactions carried out by USB and how packet serialization and de-serialization, error checking is done.

Chapter 1

OVERVIEW

1.1 INTRODUCTION

The Universal Serial Bus is a standard communication interface developed for interconnection between computer and peripheral devices. The peripherals include pointing devices, keyboard, printers, digital cameras, etc. The bus supports the plug and play feature. This implies that there is no need to open up the PC while adding a new device and allowing the software to get installed automatically. It provides low cost, bi-directional and high speed data transfer. Using USB 2.0 specification, data rate of 480Mbps transmission can be carried out. It is a point to point interface. Multiple devices can be attached from host to hub. The main aim of the USB protocol was to reduce the number of ports for PC connectivity.

1.2 USB BACKGROUND

USB was developed by a group of seven companies namely COMPAQ, IBM, INTEL, MISCROSOFT, DEC, NEC and NORTEL in the year 1994. Though the USB 1.0 specification was developed in January 1996 which produced data rates of 1.5Mbit/s (LOW SPEED) and 12 Mbit/s (HIGH SPEED) ^[3], the widely used version was 1.1 released in September 1998.

The later version, USB 2.0 was released in April 2000. The companied Hewlett-Packard, Alcatel Lucent, Philips and above mentioned firms took a joint initiative so as to generate a high data transfer rate of 480Mbit/s.

A supplement release of USB-On-The-Go specification was carried out in December 2006. This specification enabled two USB devices to communicate with each other without the need of a separate USB host.

The latest release is USB 3.0, released in November 2008 and had the objectives to increase the data rate and reduce the power consumption as well as to increase the output power.

1.3 PROBLEM STATEMENT

The USB high speed cores have already been implemented by many companies. But these companies protect their secrets and patents due the efforts and time spend in the development of the cores. Usually the user download the code from the net for trial simulation and if they are satisfied with the code, then a license fee is paid to purchase a key which allow us the programming of the device. But the problem comes in the affordability of the users as the cores are expensive to purchase.

The rise of FPGA boards in the market and its ease of availability in the labs can help to develop our own USB host IP core. The FPGA have their own USB ports which will communicate only if the cores are run on the FPGA.

The users have a very little idea the Packet forms used in the communication between the host and device. Hence we need to implement the hardware design of the Host that implements the USB protocol.

We expect our Final Year Project will help to develop a student research based; non-commercial IP core that implements a USB host and this host can be used by further generations to develop more advanced cores.

1.4 IMPLEMENTATION TOOLS

Several developmental tools were used for the implementation of our project. This includes generating Test-bench waveform, RTL simulations etc. and design summary.

We used Xilinx ISE (integrated software environment) 10.1 software for designing out circuit using Verilog code and Developing the Test-bench and schematics of the modules.

This software allows us to take our design from design entry through Xilinx device programming. The ISE project navigator processes our design through various steps in the ISE design flow.

The following are the steps used

- Design Entry
- Synthesis
- Implementation
- Simulation and verification
- Device Configuration

The Test-bench waveform containing the signals can be used to simulate the modules used in our project in the Xilinx ISE simulator.

The hardware implementation was done using Spartan 3E starter kit. This provides a powerful and highly advanced self-contained development platform for designs targeting the Spartan 3e FPGA from Xilinx. Features like Xilinx Platform Flash, USB end, JTAG parallel programming interfaces are also found on this board.

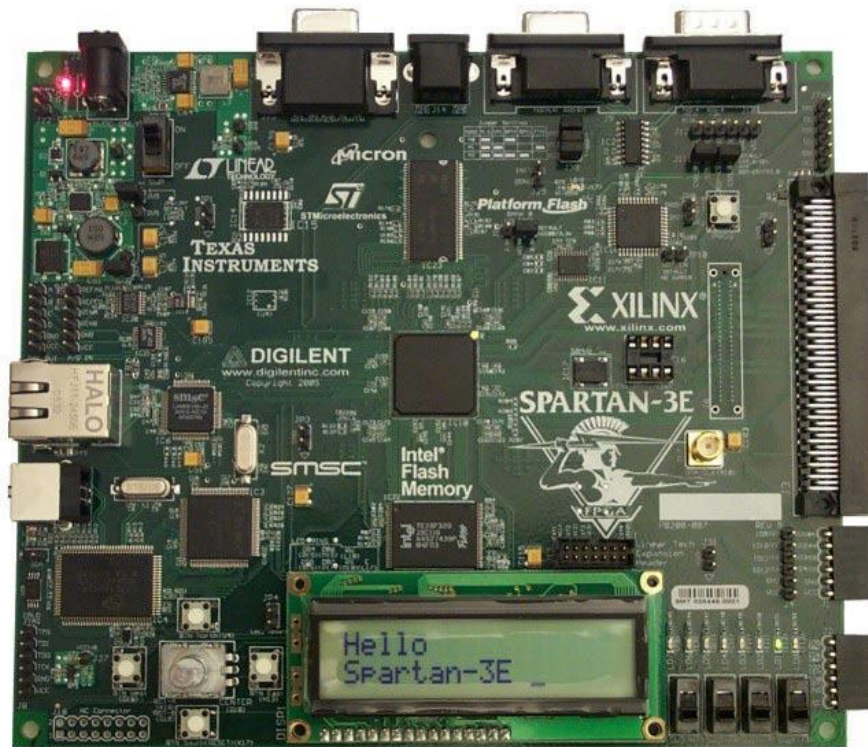


Figure 1.1 Spartan 3E FPGA board^[10]

Spartan 3E Starter Board details

IC	Xilinx Spartan 3E FPGA (500K gates)
Connector	100-pin Hirose FX2 connector Three 6-pin Pmod connectors DB15HD VGA PS/2 keyboard Two DB9 RS-232 connectors RJ-45 Ethernet 16-pin header for optional LCD modules SMA connector
Programming:	JTAG programming through on-board USB2 port; JTAG & SPI Flash programming with parallel or USB JTAG Cable; numerous additional configuration options

Chapter 2

SPECIFICATIONS

2.1 SYSTEM DESCRIPTION

A USB system has the following intrinsic areas

- USB interconnect
- USB devices
- USB host

USB interconnect: It describes the manner in which the USB devices are connected and communicates with the attached host. This includes

- Bus topology: The connection model between Host and USB devices
- Inter-layer Relationships: In terms of stack capability, the USB tasks are performed at each layer in the system.
- Data Flow Models: It is the manner in which data traverses in the system over the USB between producers and consumers.
- USB Schedule: The USB provides a shared interconnect. The access to the interconnect are scheduled so as to support isochronous data transfers and to eliminate arbitration overhead.

Bus topology: It follows a tier-star topology in which the Hub is at the center of each star and there is a point to point connection between host and hub /function or hub connected to hub/function. The maximum number of tiers allowed is 7 including the root hub.

Each compound device covers two tiers; hence it can't be enabled if it is attached at tier-7.

This is because only functions can be enabled at tier-7.

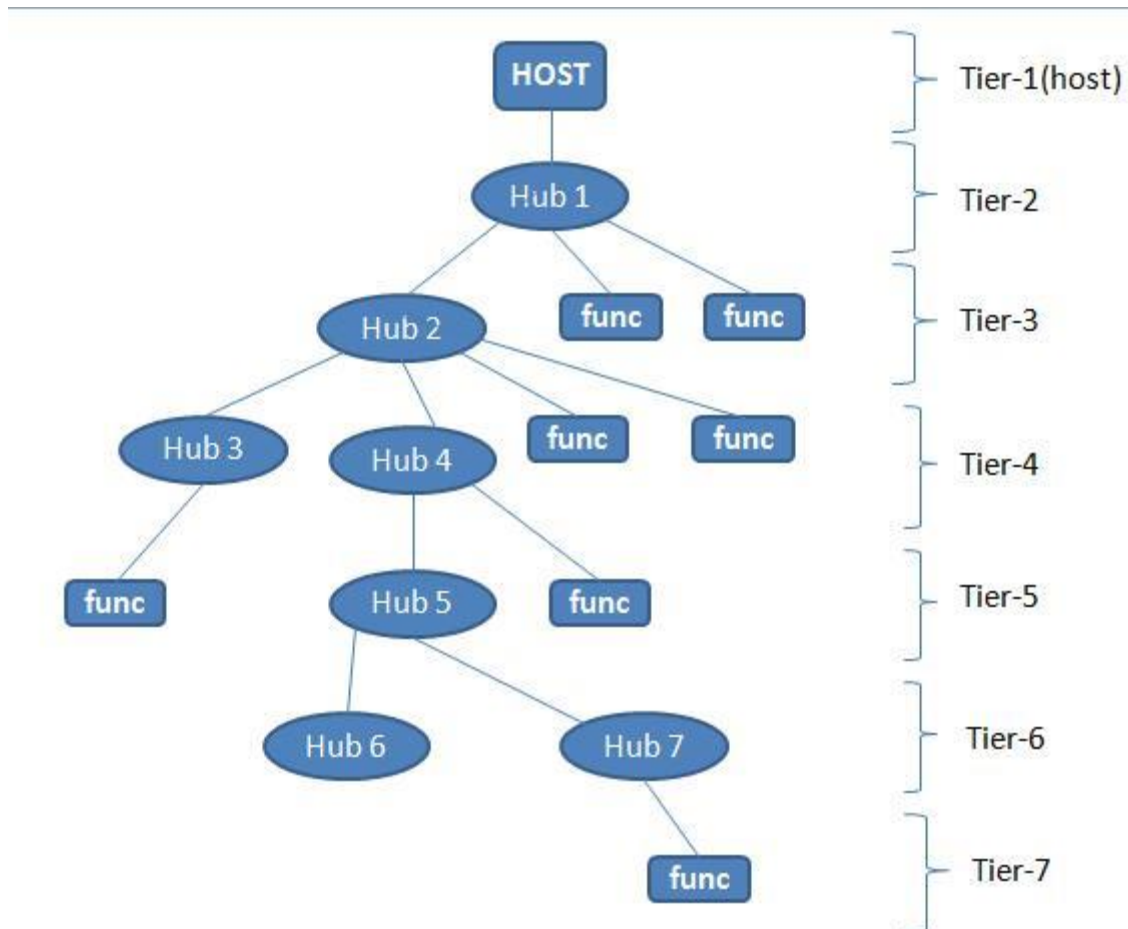


Figure 2.1 USB Bus Topology

USB Devices: The USB devices can be either a hub or a function. A hub provides additional attachment points to the USB. A USB device may comprise of many logical sub-devices which are known as device functions. Up to 127 devices can be attached to a single host controller.

USB Host: In any USB system there is only one Host. A root hub is integrated within the host system to provide one or more attachment points. The USB interface to the Host is known as Host Controller, which may be implemented in a combination of hardware, software or firmware.

2.2 COMMUNICATION FLOW

The USB cable serves as the communication path between the attached USB device and the host. The host has 3 distinct layers as shown in the figure below. Similarly the device attached to the host has 3 layers. The physical communication takes place through the lower most layers i.e. the USB wire. There is Logical communication between the host and the device above the physical layer

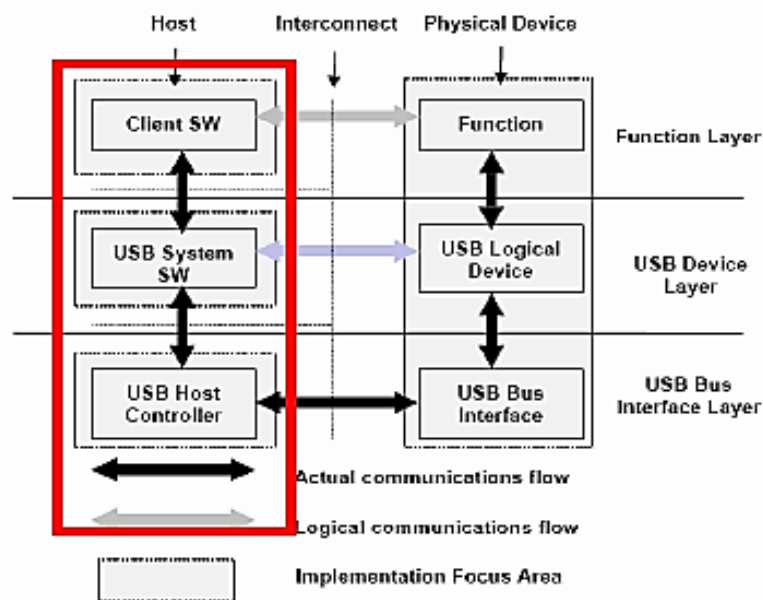


Figure 2.2 Host-Device communication model

For the purpose of our project we tried to implement few portions of the host side. The further discussions will be on the layers to the left of the above figure. The dark horizontal arrow shows the physical connectivity between the host and the device whereas the light horizontal arrow shows the logical inter-layer connectivity among the host and device. The vertical arrow shows the intra host and intra device layers connectivity.

The figure in the next page gives more detail about the different layers of a USB host.

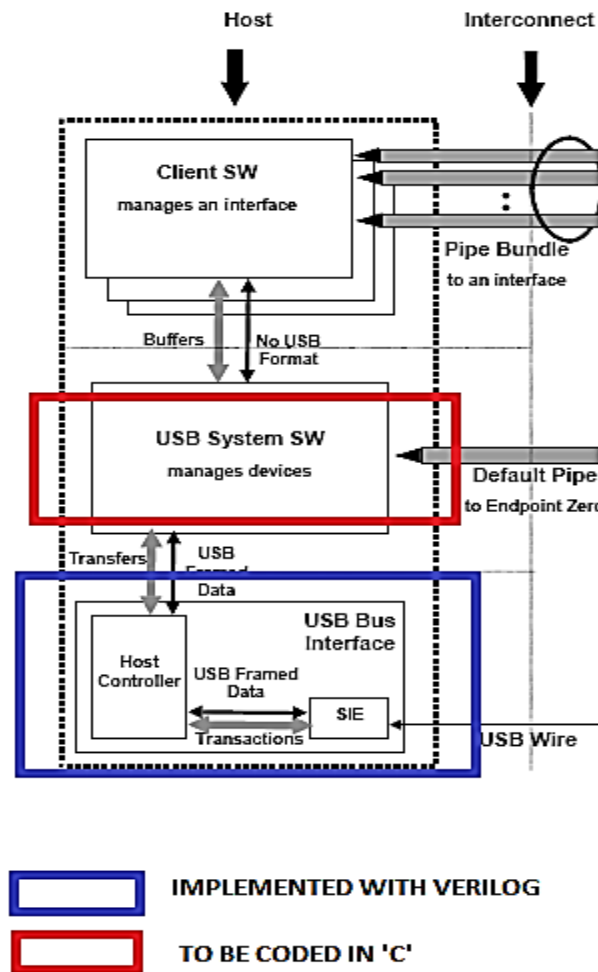


Figure 2.3 Implementation Areas^[3]

We mostly focused on the lowermost physical layer (in blue highlights) which accepts data from the level above and send in the bits to the USB cable. We expect in future the Host Controller Driver code would be written in C, and the hardware would be tested. The software interface is completely dependent on the hardware implementation and need not follow any USB driver specification. The Software layer collects the details of the protocols that are implemented in the hardware layer. The Software layer is involved in managing transactions whereas the physical layer is involved in managing packets and bits. The topmost layer is the Client Software Layer, typically in 'C' code, interacts with the USB device using high level functions like write_USB() and read_USB(). The Client layer was not implemented as well in our Project.

Physical Layer

The Physical Layer, also known as USB Bus interface Layer, handles the communication of raw data bits upon the USB wire. This is the Lowermost Layer. It comprises of SIE (Serial Interface Engine) and the HC (Host Controller). The data coming out of the USB host, passes through the HC first and then through SIE.

SIE (Serial Interface Engine): It performs many functions like serialization and deserialization of all transmissions, decoding and encoding of the signals. Verification and generation of CRCs (Cyclic Redundancy Check) and detection of PIDs (Packet Identifiers) and some special signals

HC (Host Controller): It initiates the transactions and controls the access to USB. It generates frames by dividing the time intervals and issues a SOF (Start of Frame) packet at every frame interval. Along with this, the **HC** sends requests for data to the device and from the device. It carries out error handling as well.

2.3 HOST-DEVICE COMMUNICATION

Host to device communication is carried on the basis of logical channels known as Pipes. It is a connection from host controller to logical entity found on a device. This logical entity is what we call as Endpoints. USB Device can have up to 32 endpoints. An Endpoint is inbuilt to a USB device by the designers and hence exists permanently whereas the Pipes may be opened and closed.

The figure in the next page describes pipes, endpoints and communication flow in details.

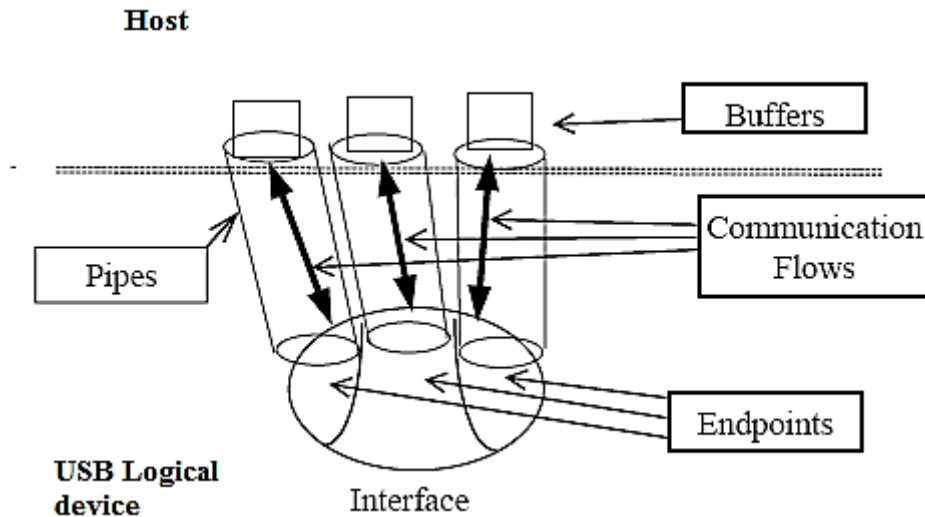


Figure 2.4 Endpoint Communication Flow^[3]

2.3.1 Device Endpoints

Endpoints are the terminal points of a communication flow between the host and a device. Each device consists of independent endpoints and has a unique address assigned to it by the system at the time of device attachment. At the time of design each endpoint is given a unique device specific identifier, known as endpoint number. Each endpoint has a device-specific direction of data flow. Each endpoint is uniquely referenced by its device address, endpoint number and direction of data flow. The data flow is simplex in nature: either input (device to host) or output (host to device).

Endpoint Characteristics:

- Bandwidth requirement
- Endpoint number
- Maximum Payload Size
- Transfer type
- Data transfer direction
- Bus access frequency

2.3.2 Pipes

USB pipe brings out connectivity between the endpoint on a device and software on the host. It represents the ability to move data between the host software through the memory buffer and device endpoint. Pipes communication exists in two types.

Stream pipe: The data packet moving in this mode has no USB-defined structure. Stream pipe is unidirectional. It supports isochronous, bulk and interrupt transfers. The data coming in from one end and coming out of the other end has the same order. The data moves through this pipe in sequential order i.e. FIFO (First in First out).

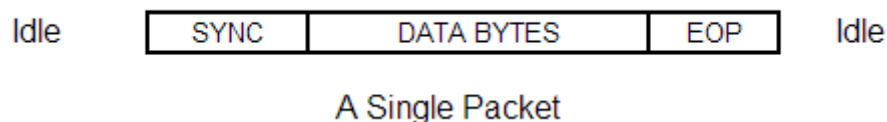
Message pipe: The data packet moving in this mode has some USB-defined structure. Message pipe is bi-directional and it supports control transfer. It requires single endpoint number during the communication in both directions. The default control pipe always represents a message pipe.

2.3.3 Descriptor table

A descriptor is a data structure with a specific format. Each descriptor starts with a byte-wide field which contains the total number of bytes in it followed by a byte sized field that identifies the descriptor. A device descriptor table includes the general information about a USB device. This information is applicable globally to the device and its entire device configuration. Each USB device has only one device descriptor.

2.4 PACKET DESCRIPTION

The data transmitted in USB communication are called packets. The packets are sending from host to devices via USB hubs. The packets follow a specific format. It starts with a SYNC pattern which allows the receiver bit clock to synchronize with the data, followed the data bytes of the packet (Pay Load) and concludes with EOP (End of Packet) signal. The data is NRZI (Non Return to Zero Inverted) encoded and in order to ensure sufficiently frequent transitions, it is bit-stuffed. Before and after the packet transmission, the bus remains in the idle state.



2.4.1 Common USB packet fields

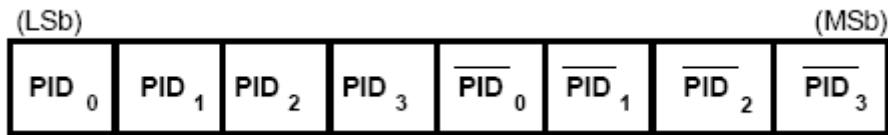
On the USB data bus, the LSBit is transmitted first for any data transmission. The USB packet consists of the following fields.

- **SYNC**

All packets begin with the SYNC field which is 8 bits long at low and full speed or 32 bits long for high speed. It is used to synchronize the receiver clock with that of the transmitter clock. The last 2 bits indicates the end of SYNC field and start of PID field.

- **PID**

PID or Packet identifier comes immediately after the SYNC field. It consists of 4 bit packet type field followed by a 4 bit check field. This check field is generated by doing 1's complement of the packet type field. Error occurs if the Check bits are not complement of the Packet type bits.



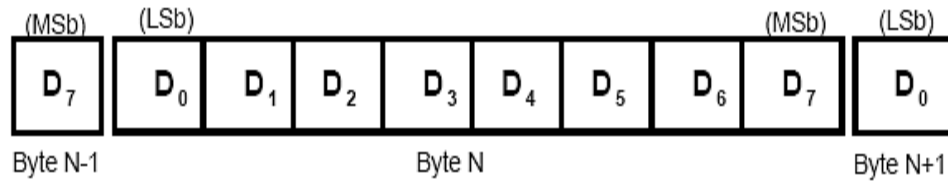
The PID indicates the type of packet i.e. format of the packet and type of error detection applied to the packet. According to USB 2.0 specifications there are 17 different PID values are defined. These are mentioned below in the table.

PID Type	PID Name	PID<3:0>*	Description
Token	OUT	0001B	Address + endpoint number in host-to-function transaction
	IN	1001B	Address + endpoint number in function-to-host transaction
	SOF	0101B	Start-of-Frame marker and frame number
	SETUP	1101B	Address + endpoint number in host-to-function transaction for SETUP to a control pipe
Data	DATA0	0011B	Data packet PID even
	DATA1	1011B	Data packet PID odd
	DATA2	0111B	Data packet PID high-speed, high bandwidth isochronous transaction in a microframe
	MDATA	1111B	Data packet PID high-speed for split and high bandwidth isochronous transactions
Handshake	ACK	0010B	Receiver accepts error-free data packet
	NAK	1010B	Receiving device cannot accept data or transmitting device cannot send data
	STALL	1110B	Endpoint is halted or a control pipe request is not supported
	NYET	0110B	No response yet from receiver
Special	PRE	1100B	(Token) Host-issued preamble. Enables downstream bus traffic to low-speed devices.
	ERR	1100B	(Handshake) Split Transaction Error Handshake (reuses PRE value)
	SPLIT	1000B	(Token) High-speed Split Transaction Token
	PING	0100B	(Token) High-speed flow control probe for a bulk/control endpoint
	Reserved	0000B	Reserved PID

Table 2.1 Packet Identifier (PID) description table

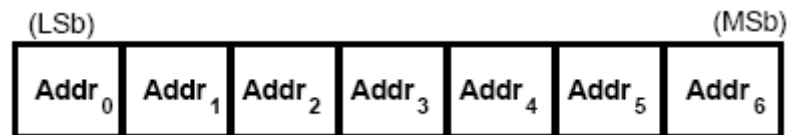
- **DATA**

The data field must be an integral number of bytes and ranges from 0-1024 bytes. The shifting of data begins from the LSBit. The size of packet varies according to the transfer type used.



- **ADDR**

The address field (ADDR) specifies the destination of the data packet based on the value of token PID. It is 7 bits in length hence allowing a maximum of 127 devices to be used concurrently. Address 0 is considered invalid because any device which has not been assigned any address must respond to the packets sent at address 0.



- **ENDP**

The Endpoint field (ENDP) allows flexible addressing of the functions which require more than one endpoint. It is 4 bits long thus providing 16 possible endpoints. A function can support a maximum of 16 IN and 16 OUT endpoints.

- **Frame Number Field**

This field is 11 bits long and is incremented by the host on per-frame basis. Its maximum value is 7FFH and it is sent in SOF tokens at the beginning of each frame/micro frame.

- **CRC**

A Cyclic Redundancy Check (CRC) is a value calculated from a number of data bytes to generate a unique value which is transmitted along with the data bytes, and then it is used to validate the correct reception of the data. Token packets have CRC of 5 bits while data packets have a CRC of 16 bits.

- **EOP**

The packet field concludes with End of Packet (EOP). It is signaled by SE0 (Single Ended Zero) for approximately 2 bit times and then followed by a J for 1 bit time.

2.4.2 USB Packet Types

There are four different packet formats depending on the type of PID used at the beginning of the packet.

- **Token Packet**

Token packet indicates the type of transaction to be followed and identify the targeted endpoints. There are 3 types of Token packet IN, OUT and SETUP.

IN packet indicates the USB device that the Host wants to read information.

OUT packet indicates the USB device that the Host wants to send information

SETUP packet is used to begin control transfers.

SYNC	PID	ADDR	ENDP	CRC5	EOP
	8 bits	7 bits	4 bits	5 bits	

Table 2.2 Token packet format

- **Data Packet**

Data PIDs DATA0 and DATA1 are used transmitting data of size up to 1024 bytes. These PIDs are used in Low and Full speed links. Data PIDs DATA2 and MDATA are used in High speed mode.

SYNC	PID	DATA	CRC16	EOP
	8 bits	(0-1024)*8 bits	16 bits	

Table 2.3 Data packet format

Max data payload for low speed: 8 bytes

Max data payload for full speed: 1023 bytes

Max data payload for high speed: 1024 bytes

- **Handshake packet**

Handshake packets reports the status of data transaction and returns values indicating successful data reception or rejection, flow control and halt conditions. There are 4 types of handshake packets: ACK, NACK, NYET, and STALL.

SYNC	PID	EOP
	8 bits	

Table 2.4 Handshake packet format

- **SOF**

SOF (Start of Frame) indicates the start of a new frame. It is send every 1ms on full speed links.

SYNC	PID	Frame No.	CRC5	EOP
	8 bits	11 bits	5 bits	

Table 2.5 Start of Frame (SOF) packet format

Chapter 3

DESIGN

ARCHITECTURE

3.1 GENERAL ARCHITECTURE

The following diagram gives the overall view of the core architecture. The host interface acts like a bridge between the internal data memory and control registers with the function controller. The control registers and the data memory interface to the Protocol Layer (PL). The PL interfaces to the UTMI block. Then the UTMI interface to the PHY. Each of the blocks has been described below.

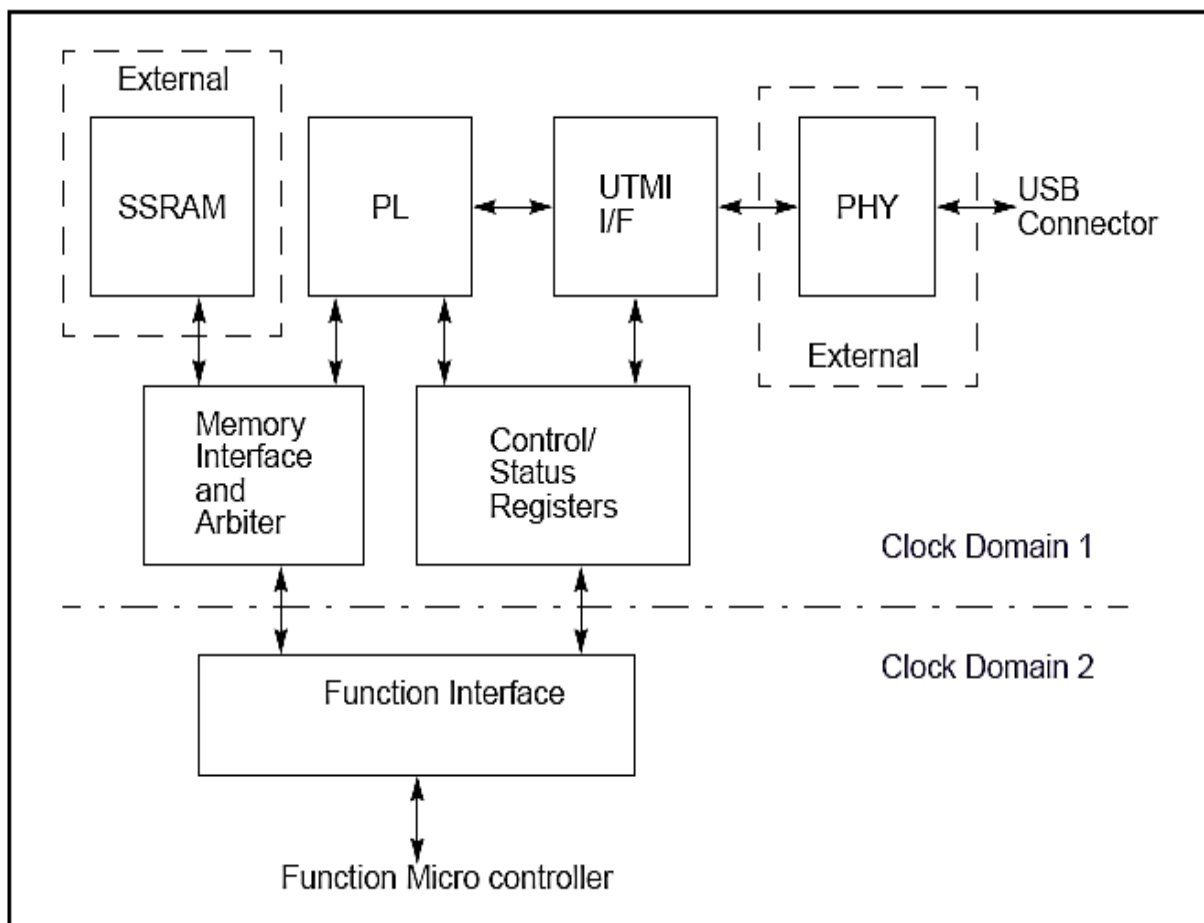


Figure 3.1 USB Core architecture

3.1.1 Clocks

The USB core consists of 2 clock domains. The UTMI runs on the clock provided by PHY which is 60 Mhz. The real clock frequency is dependent on the modes that are operated (High speed/Full speed). The UTMI synchronizes to the rest of the core.

Other blocks run on the clock from host interface. Due to latency requirement, the host interface must run at least 60 MHz. The main aim is that the minimum required frequency of the USB host interface should be at least 100 MHz.

3.1.2 Host Interface

It provides a consistent interface between the internal functions of the USB core and the function-defined host or micro-controller. The maximum theoretical throughput of the USB is 480 Mb/s or 60MB/s. On a bus size of 32 bits, 4 bytes are transferred in one cycle. Hence the minimum bandwidth required now for the host is 15 Mwords/s. (1 word = 4 bytes)

3.1.3 Memory Interface and Arbiter

This block arbitrates between the USB core and the host interface so as to access the memory. This block uses a standard single port SRAM (synchronous). This block also performs dataflow management and flow control.

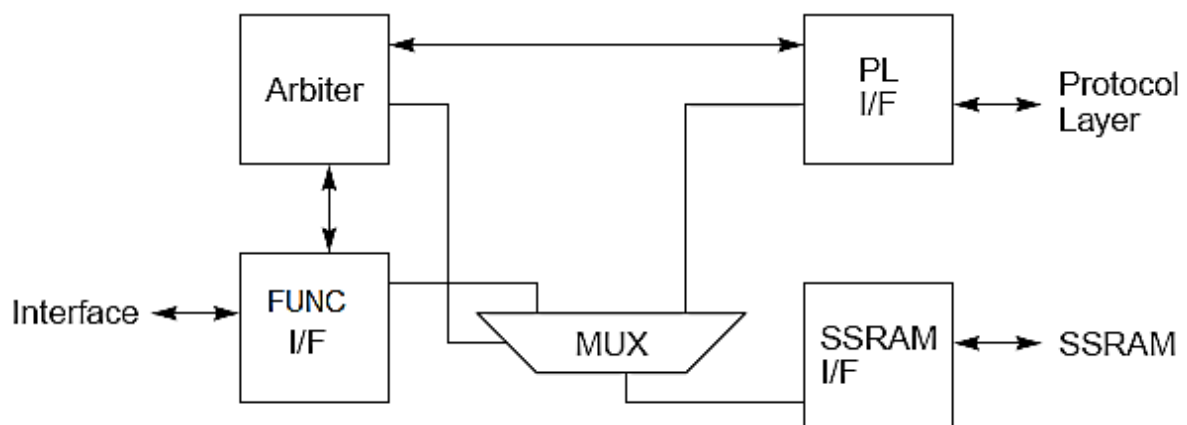


Figure 3.2 Memory Interface and Arbiter

3.1.4 SSRAM

The SSRAM (Synchronous Static Random Access Memory) block is used for buffering the input and the output data.

3.2 UTMI INTRODUCTION

The UTMI stands for **USB 2.0 Macrocell Transceiver Interface**. This block takes care of low level USB protocols and signaling. The features which are included in signaling are data serialization and deserialization, bit stuffing and unstuffing, clock recovery and synchronization. The main focus of this block is to shift the compatibility of the clock domain of the data from USB to that of the logic implemented out FPGA. The FPGAs operate at the frequency range of 30-60 MHz, whereas the USB signaling runs at hundreds of MHz clock speed.

Some of the key features of the UTMI are given below:

- Enables multiple IP source for the SIE (Serial Interface Engine) part
- Supports 3 speed modes, HS /FS/LS serial data transmission
- Uses 8 bit parallel interfacing to send and receive data
- SYNC/EOP field generation and checking
- Supports Dual Speed mode devices
- Data recovery and Clock recovery from serial stream
- Bit stuff error reorganization
- Hold register to receive and transmit data
- Resume and Suspend signaling logic
- Switching between FS and HS modes
- Single Parallel clock with on-chip PLL generates high-speed serial clocks

The Signal SuspendM brings the USB 2.0 into low power mode.

The functional block level diagram (in section 3.4) was followed in the design of the UTMI.

These blocks can be discussed only after having a proper understanding of the System and Data Interface Signals used in UTMI design.

3.3 UTMI SIGNAL DESCRIPTION

3.3.1 System Interface Signals

NAME	DIRECTION	ACTIVE LEVEL	DESCRIPTION
CLK	Output	Rising edge	The output is used for clocking the receive and transmit parallel data
Reset	Input	High	It resets all the state machines
Xcvr Select	Input	N/A	Transceiver Select: It selects the FS/HS mode transceivers 0: HS transceiver enabled 1: FS transceiver select
Term Select	Input	N/A	Termination Select: It selects between FS/HS terminations 0: HS termination enabled 1: FS termination enabled
SuspendM	Input	Low	It places the Macrocell in that mode which draws minimal power from the sources. 0: Macrocell drawing Suspend current 1: Macrocell not drawing suspend current
LineState (0-1)	Output	N/A	Line State: It shows the current state of the receivers. DM DP Description 0 0 : SE0 0 1 : 'J' State

			1 0 : 'K' State 1 1 : SE1
OpMode (0-1)	Input	N/A	Operation Mode: It selects the operational modes [1] [0] Description 0 0 0: Normal Operation 0 1 1: Non-Driving 1 0 2: Disable Bit Stuffing and NRZI encoding 1 1 3: Reserved

3.3.2 USB interface signals

NAME	DIRECTION	ACTIVE LEVEL	DESCRIPTION
DP	Bidirectional	N/A	USB data pin: Data+
DM	Bidirectional	N/A	USB data pin: Data-

3.3.3 Data interface signals (Transmit)

NAME	DIRECTION	ACTIVE LEVEL	DESCRIPTION
DataIn (0-7)	Input	N/A	8-bit parallel data input bus.
TxValid	Input	N/A	Transmit Valid: It Indicates that the DataIn bus is valid
TxReady	Output	High	Transmit data ready: If TxValid and TxReady are high then UTMI loads DataIn to Tx register.

3.3.4 Data interface signals (Receive)

NAME	DIRECTION	ACTIVE LEVEL	DESCRIPTION
DataOut(0-7)	Output	N/A	8-bit parallel data output bus.
RxValid	Output	N/A	Receive Valid: It Indicates that the DataOut bus is valid
RxActive	Output	High	Receive Active: Indicates the detection of SYNC and is active
RxError	Output	High	Receive Error

3.4 UTMI BLOCK LEVEL DESCRIPTION

This section contains the block diagram of UTMI. Each high level functionality block, which exists in the Macrocell, is described.

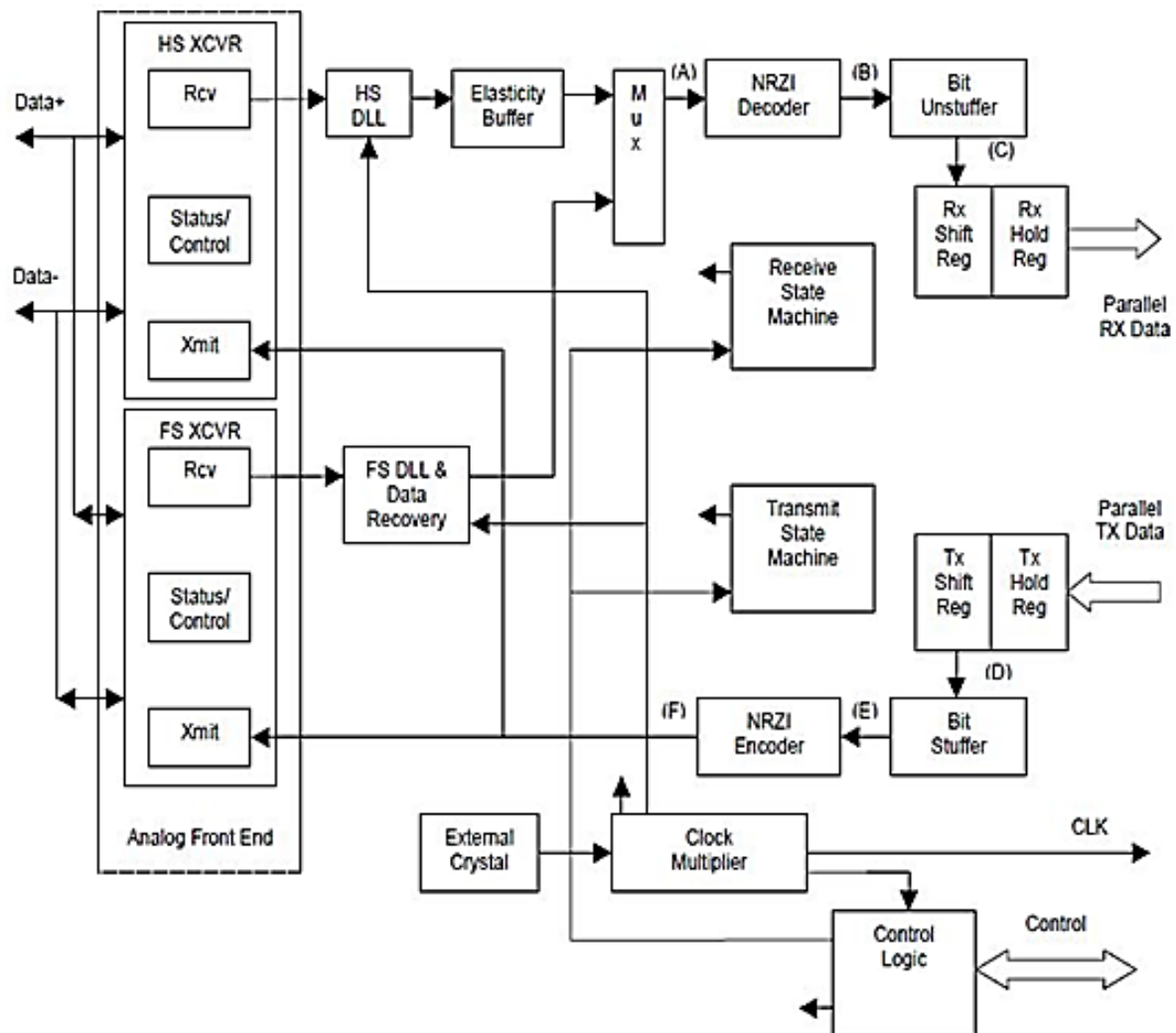


Figure 3.3 UTMI block level diagram

3.4.1 Clock Multiplier

This module produces the proper internal clock for UTMI and **CLK** as output signal. The data transfer signals are in synchronization with the **CLK** signal. The CLK frequency error should be less than 10%. The CLK relation with the data and control signals can be shown with the following Receive and Transmit data.

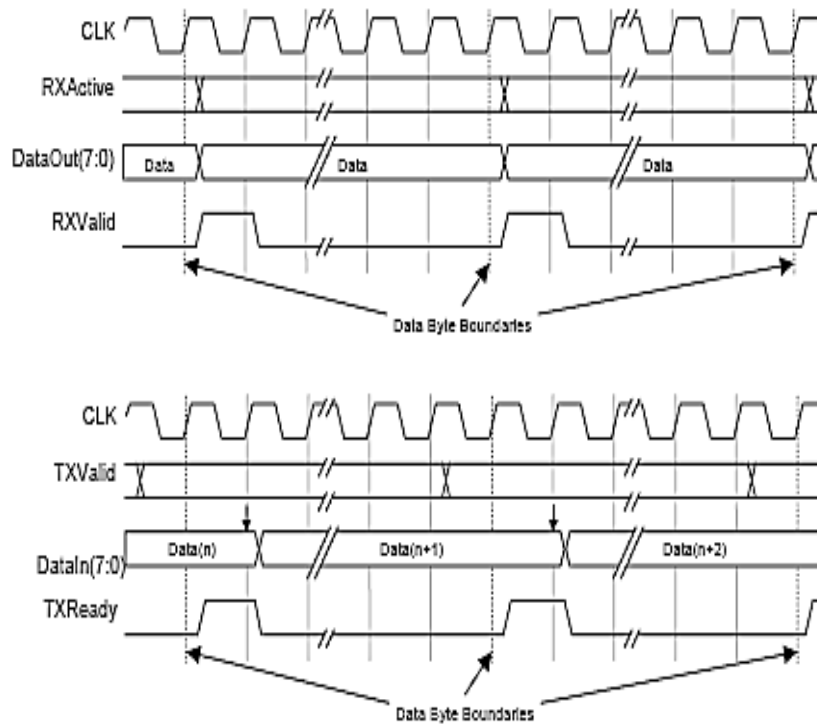


Figure 3.4 FS CLK relations to Receive/Transmit data and Control signals

3.4.2 HS DLL (HS Delay Line Phase Locked Loop)

This block extracts the clock and data from the received data over the USB interface so as to be received by Receive Deserializer. The output from the DLL stays in synch with the local clock.

3.4.3 Elasticity Buffer

This block is used to compensate any difference between the receiving and transmitting clocks. If underflow or overflow condition is detected then it is reported with the **RxError** signal.

3.4.4 Mux

XcvrSelect input decides the state of Mux. Accordingly it routes the data from HS or FS receivers to the shared receive logic.

3.4.5 NRZI Decoder

The received data on DP, DM lines are NRZI decoded. The NRZI Decoder simply XOR the present bit with the provisionally received bit. During the NRZI decoding, receive state machine remains in RX wait state

3.4.6 Bit Unstuff Logic

This logic can operates on both FS or HS data rates. The Bit Unstuffer checks each bit of the data stream and if a zero is detected after six consecutive '1's that zero bit is deleted. In FS mode, if error is detected then **RxError** signal is asserted. Whereas in HS mode, an EOP signal is generated using the bit stuff errors. Hence **RxError** signal is not asserted in HS mode.

3.4.7 Rx Shift/Hold Register

This block deserializes the received data , recovered by the DLL and transmit 8-bit parallel data to the bus interface. This module is primarily an 8-bit shift register which does serial to parallel conversion and buffers the deserialized data byte.

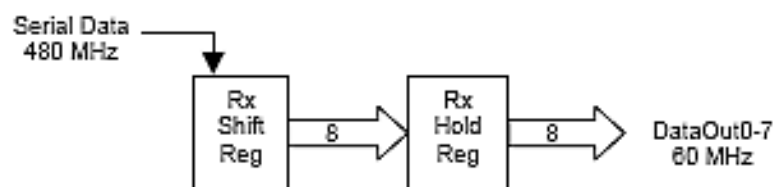


Figure 3.5 Rx Shift/Hold Register

3.4.8 Receive State Machine

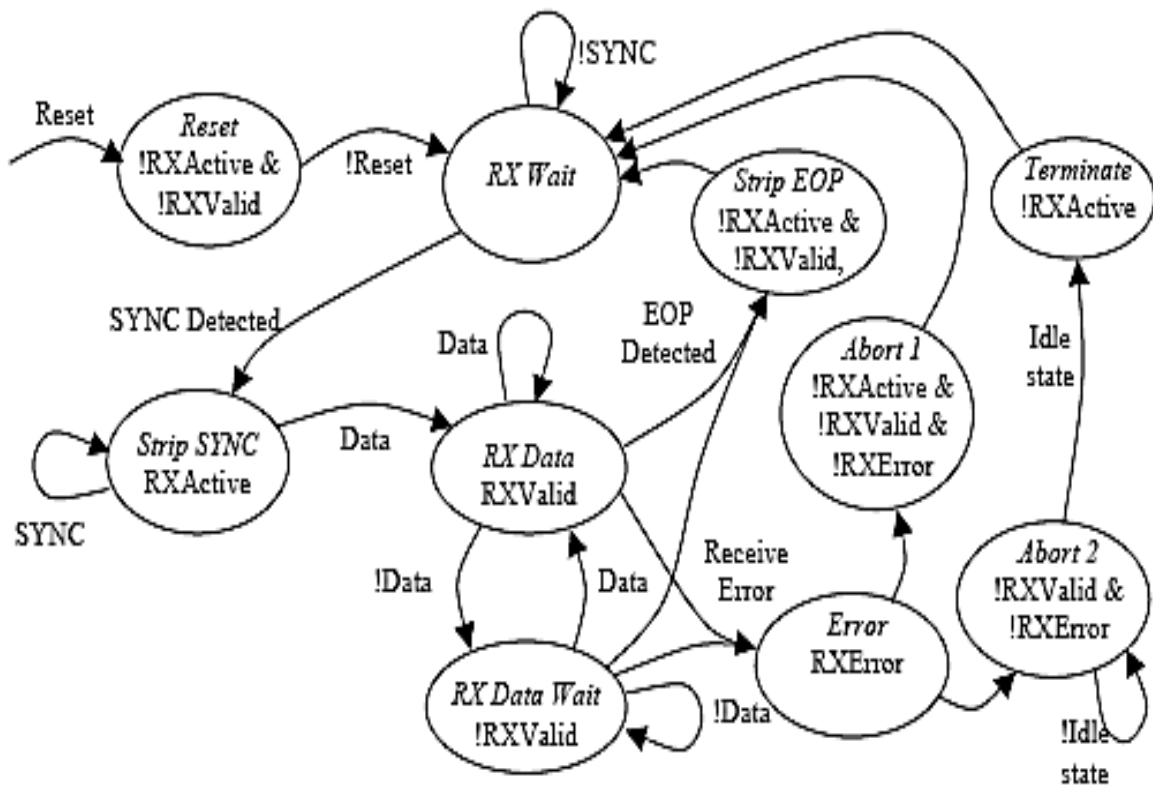


Figure 3.6 Receive state machine

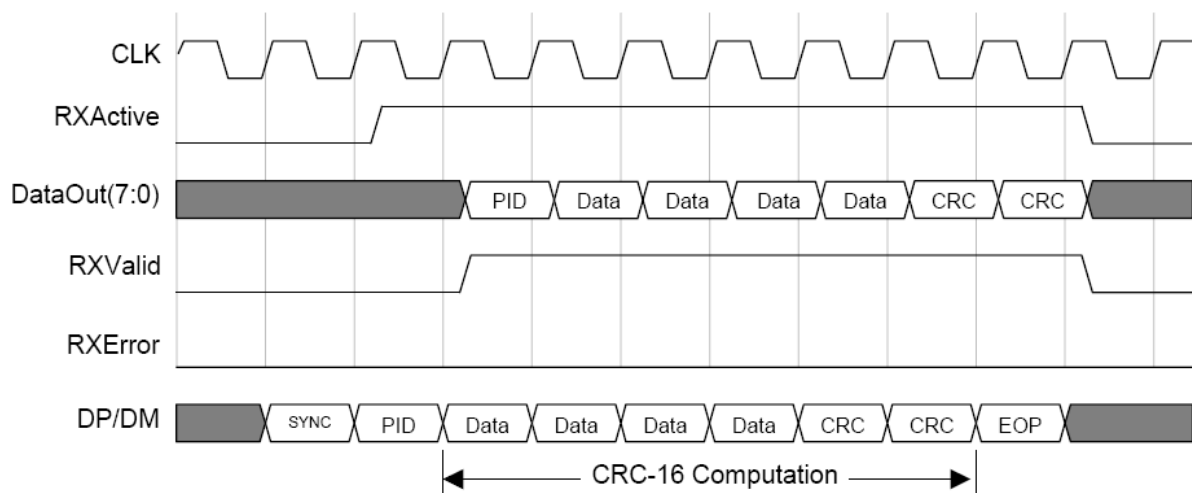


Figure 3.7 Receive timing cycle

When the Reset signal is negated, the Receive State Machine enters the Wait state as shown. There it starts looking for a SYNC pattern. When it has detected a SYNC pattern, the state machine will enter the Strip SYNC state and assert RXActive. After receiving 8 bits of valid serial data, the state machine enters the RX Data state where the data is loaded into the RX Holding Register on the rising edge of CLK and RXValid is asserted. Stuffed bits are then stripped from the data stream. Each time 8 stuffed bits are accumulated, state machine will enter the RX Data Wait state, negating the RXValid.

When the EOP signal is detected, the state machine will enter the Strip EOP state and negate RXActive and RXValid. After the EOP has been stripped the RX State Machine will re-enter the Wait state and begin looking for the next packet.

If a Receive Error is detected in the process, the state machine enters Error State and RXError is asserted. Then state machine enters either the Abort 1 State where RXActive, RXValid, and RXError signals are negated, or the Abort 2 State where only RXValid, and RXError signals are negated

3.4.9 NRZI Encoder

The transmitted data on DP, DM lines are NRZI encoded. During the NRZI encoding, transmit state machine remains in TX wait state. When bit '1' is received in the serial data it is negated and transmitted to DP, DM lines. Whereas when bit '0' is encountered, it is directly transmitted on to the DP, DM lines.

3.4.10 Bitstuff Logic

Bit stuffing is carried out so as to ensure adequate signal transition when sending data. A '0' is inserted after every 6 consecutive 1s before the data is NRZI encoded. This is done to enforce a transition in the NRZI data stream. SYNC pattern enables Bit stuffing.

After every 8 bits stuffed into the data stream, **TxReady** gets negated for one byte time so as to hold up the data on to the **DataIn** bus.

3.4.11 Tx Shift/Hold Register

The module receives 8bit parallel data and buffers in the Tx Hold Reg and then sends to 8-bit shift register. This block primarily converts parallel data to serial data

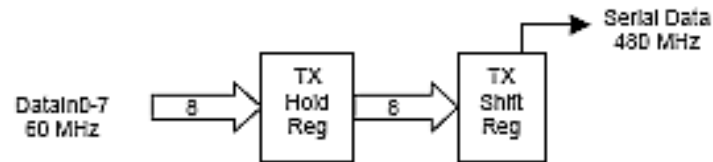


Figure 3.8 Tx Shift/Hold Register

3.4.12 Transmit State Machine

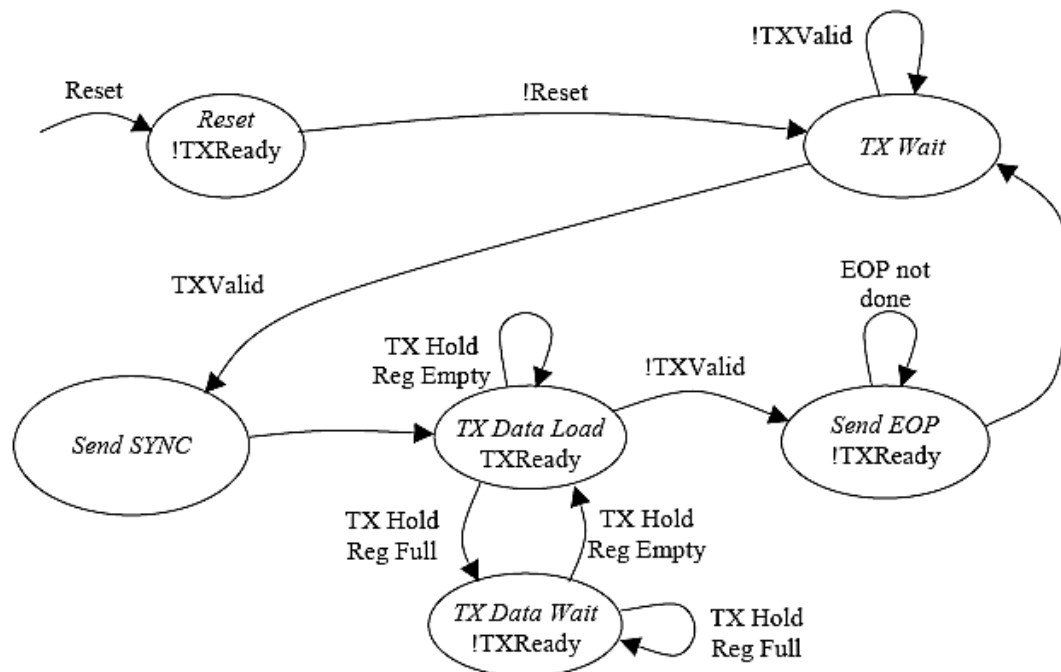


Figure 3.9 Transmit state machine

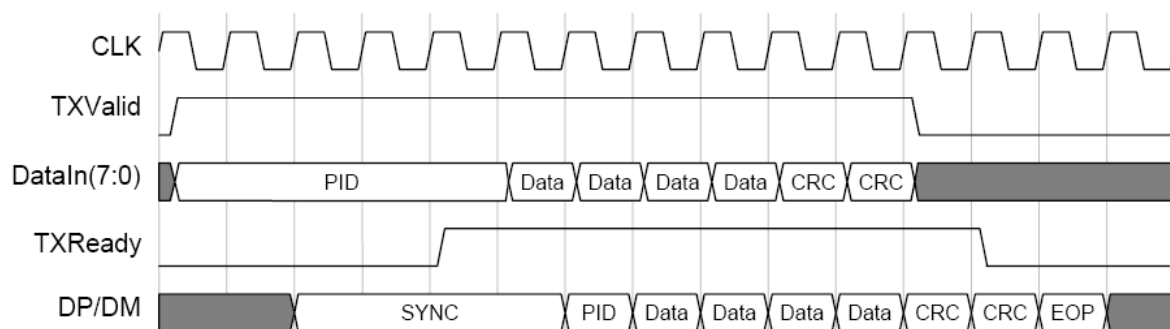


Figure 3.10 Transmit timing cycle

When Reset signal is negated the, transmit state machine enters the TX Wait state. In the TX Wait state, the TX state machine looks for the assertion of TXValid and when it is detected, the state machine will enter the Send SYNC state and begin transmission of the SYNC pattern. When the transmitter is ready for the first byte of the packet (PID), it enters the TX Data Load state, TXReady is asserted and the TX Holding Register is loaded. The state machine may enter the TX Data Wait state on completion of the SYNC pattern transmission.

The state machine remains in the TX Data Wait state until the TX Data Holding register is available for some more data. In the TX Data Load state, the state machine loads the Transmit Holding register. The state machine will remain in TX Data Load state as long as the transmit state machine can empty the TX Holding Register before the next rising edge of CLK.

When TXValid is negated, the TX state machine enters the Send EOP state where it sends the EOP. During the EOP transmission, TXReady signal is negated and the state machine will remain in the Send EOP state. After the EOP is transmitted the Transmit State Machine returns to the TX Wait state, looking for more data

3.4.13 USB HS/FS XCVR

Transmit driver: On enabling, the data from the transmitting data path will be driven into DP/DM Signal lines. The HS/FS transmit driver becomes active only when it is asserted. The **XcvrSelect** input is in HS/FS transceiver enabled mode.

Receive Buffer: On enabling, the received HS/FS data is multiplexed through the data path of the receiver to the shift and hold register of the receiver. The USB receive buffers gets active when the **XcvrSelect** input is in the HS transceiver enabled mode.

3.5 PROTOCOL LAYER

The protocol layer carries out all the USB data input/output and communications in the USB core.

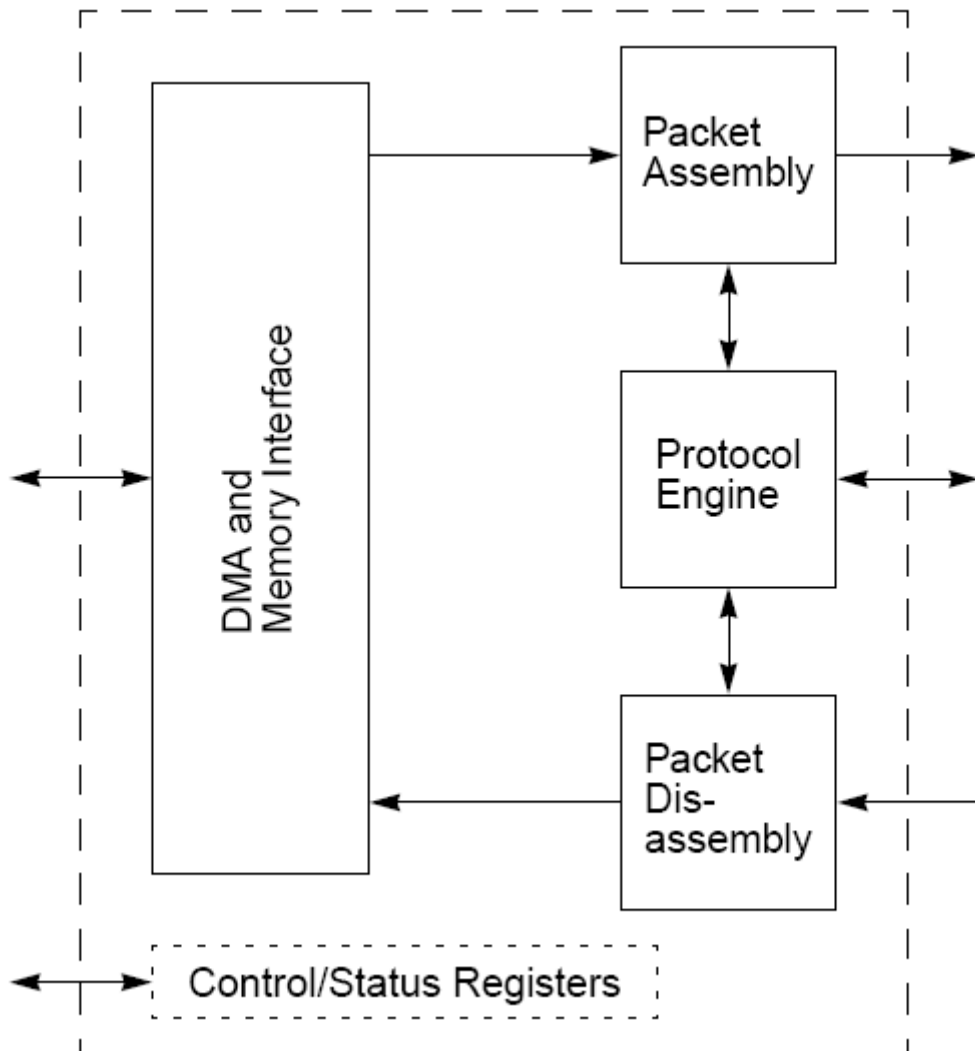


Figure 3.11 Protocol Layer overview

DMA and Memory Interface

DMA interfaces with the data memory. This provides DMA block transfer and random memory access.

Protocol Engine

Protocol Engine performs USB's standard handshake and control correspondence. These are SOF tokens, ACK, NYET, NACK, replying to PING tokens.

Packet Assembly

This block collects the packets and puts them into the UTMI. First, it assembles the header, then it inserts proper PID and Checksum, finally data field is added (if requested).

Packet Disassembly

This block recognizes the incoming packets and sends the decoded data to the proper block. The decoding process includes extraction of PID and sequence number. Along with that recognizing the header check sum is also carried out.

3.5.1 Buffer Operations

The buffer operations are carried out by buffer pointers which point to the I/O data structures in the memory. If the value of buffer pointer is 7FFFh, then it indicates that the buffer hasn't been allocated yet. The core responds with a **NAK** to the USB host if all the buffers are not allocated. Data transfer within buffer takes place in a Round Robin fashion. Buffer 0 is used first when data is sent to/from an endpoint. The function controller is notified through an interrupt when the Buffer 0 is empty/full. Buffer 0 can now be refilled/emptied by the function controller. When the second buffer is empty/full, the function controller is interrupted, and the USB core will use buffer 0 again, and so on.

A buffer can be larger than the `MAXIMUM_PAYLOAD_SIZE`. In that case, many packets can be retrieved/placed from/to a buffer. For an OUT endpoint, a buffer must always be a multiple of maximum payload size. The buffer size field should always be checked by the software. When the entire buffer has been used, the buffer size is 0. If it is not 0, then the size field indicates the number of bytes of the buffer that has not been used.

There is no such limitation in case of IN buffers. The core will always transmit the maximum possible number of bytes, which is the minimum of the payload size and remaining buffer size.

Control endpoints can receive and transmit data. Hence in this case, Buffer0 is always an OUT buffer and Buffer1 always an IN buffer.

Buffer underflow

A buffer underflow condition is specified, when either the external DMA engine or the function controller has not filled the, the internal buffer with enough data for one `MAXIMUM_PAYLOAD_SIZE` packet. In this condition, the USB core replies a NACK to the host when an IN token is received.

Buffer overflow

A buffer overflow is specified when a packet, that has been received, doesn't fit into the buffer. The packet is discarded and a NACK is sent to the host. The USB core will continue discarding the received data, and reply with NACK to each OUT token when payload size doesn't fit into the buffer.

3.5.2 DMA Operation

DMA operation allows a complete transparent movement of data from the USB core to the functions attached. Each point is associated with a pair of **`dma_req`** and **`dma_ack`** signals.

The USB core will use these signals for DMA flow control when the **`dma_en`** bit is set in the CSR register. When the buffer contains data or when the buffer is empty and needs to be filled, the **`dma_req`** signal is asserted. The DMA responds with a **`dma_ack`** signal for each word transferred.

Only Buffer 0 is used in DMA mode. The function controller is notified with an interrupt when the received packet is less than the `MAX_PL_SZ`. In addition to it, the Buffer1 is set to the local buffer address of that received packet. So as to carry out uninterrupted DMA transfers, the buffer is padded by the USB core with the `MAX_PL_SZ` bytes.

3.5.3 Control Status Register (CSR)

The CSR acts as the control word of the DMA operation. The status and configuration bits specify the type of operation mode.

BIT	ACCESS	DESCRIPTION
31:30	RO	UC_BSEL (Buffer Select) This bits must be initialized to zero (first Buffer 0 is used) 00: Buffer 0 01: Buffer 1 1x: RESERVED
29:28	RO	UC_DPD These two bits are used by the USB core to keep track of the data PIDs for high speed endpoints and for DATA0/DATA1 toggling.
27:26	RW	EP_TYPE (Endpoint Type) 00: Control Endpoint 01: IN Endpoint 10: OUT Endpoint 11: RESERVED
25:24	RW	TR_TYPE (Transfer Type) 00: Interrupt 01: Isochronous 10: Bulk 11: RESERVED
23:22	RW	EP_DIS (Disable The Endpoint Temporarily) 00: Normal Operation 01: Force the core to ignore all transfers to this endpoint 10: Force the endpoint in to HALT state 11: RESERVED
21:18	RW	EP_NO (Endpoint Number)
17	RW	LRG_OK 1 - Accepts data packet of more than MAX_PL_SZ bytes (RX only) 0 - Ignores data packet with more than MAXPL_SZ bytes (RX only)
16	RW	SML_OK 1 – Accepts data packet with less than MAX_PL_SZ bytes (RX only) 0 - Ignores data packet with less than MAXPL_SZ bytes (RX only)
15	RW	DMAEN 1: Enables external DMA interface and operation 0: No DMA operation
12:11	RW	TR_FR Number of transactions per micro frame (HS mode only)
10:0	RW	MAX_PL_SZ Maximum payload size (MaxPacketSize) in bytes

Table 3.1 Control Status Register bit definition

3.6 PROTOCOL ENGINE

The protocol engine is responsible for CRC check and generation , packet identifier decoding and verification, recognition of address and handshake evaluation and response. The protocol engine acts on a received token and analyzes the token's PID , address and endpoint number fields. The protocol engine handles USB packets and transactions based on data sequencing and state machine logic.

Protocol engine complies with High Speed USB 2.0 specification with a transfer rate of 480Mbit/s. It also meets UTMI specifications, generating control signals for UTMI transceiver interface according to the FSM states.

There are 3 types of transactions: IN, OUT and SETUP.

There are 4 data transfer types: Control, Bulk, Interrupt and Isochronous.

3.6.1 Dataflow Types

- **CONTROL TRANSFER**

Control transfers are used for initial configuration of the device by the host using Endpoint 0 OUT and Endpoint 0 IN. The data delivery is lossless. It is divided into 3 stages SETUP, DATA and STATUS.

The SETUP stage carries an 8 bytes Setup packet which defines the request and mentions the amount of data to be transferred in the DATA stage.

The DATA stage starts with a transaction which contains DATA1. The transaction switches between DATA1 and DATA0 until all the required data is transferred.

The STATUS stage contains a zero-length DATA1 packet. For DATA stage as IN, the STATUS stage is OUT and vice-versa.

Max packet size: 8 for low speed; 8, 16, 32 or 64 for full speed; 64 for high speed.

- **BULK TRANSFER**

These are designed to transfer large amount of data with error free delivery without any guarantee of bandwidth. Bulk data has a sequential form. If an IN/ OUT endpoint is defined as using Bulk transfers, then the host will transfer data to it using IN/OUT transactions respectively. Bulk transfer is not applicable for low speed.

Max packet size: 8, 16, 32 or 64 for full speed; 512 for high speed.

- **INTERRUPT TRANSFER.**

Interrupt transfer are applicable to devices that need guaranteed quick responses (e.g.: Pointing Devices and keyboards). These are regularly scheduled IN or OUT transfers.

Packet size may vary from 1-8 bytes at low speed and 1-64 for full speed or upto 1024 bytes at high speed

- **ISOCHRONOUS TRANSFER**

Isochronous transfers give a guaranteed bandwidth but without any guarantee of error-free delivery. A special feature of this transfer is, there is no handshake packet at the end. Applications such as audio data transfer follow isochronous mode of data transfer because it is important to maintain a constant flow of data.

It uses either OUT or IN transaction depending on the type of endpoint.

It is not applicable for low speed data transfer. Packet size varies upto 1023 bytes at full speed or upto 1024 at high speed.

Chapter 4

SIMULATIONS & RESULTS

4.1 UTMI SIGNALS SIMULATION

In the simulation, the **rx_data** contained Up-Counter values and **tx_data** contained Down-Counter values. For successful initiation of data transfer, the control signal **tx_valid** and **rx_valid** must be asserted. Whenever **tx_first** is negated, the **data_out** remains the previous value until the **tx_first** is asserted again.

In the simulation, the value **8'hFE** gets stored because **tx_first** become low at 1.2us. It regains data at the time 3us when **tx_ready** becomes high.

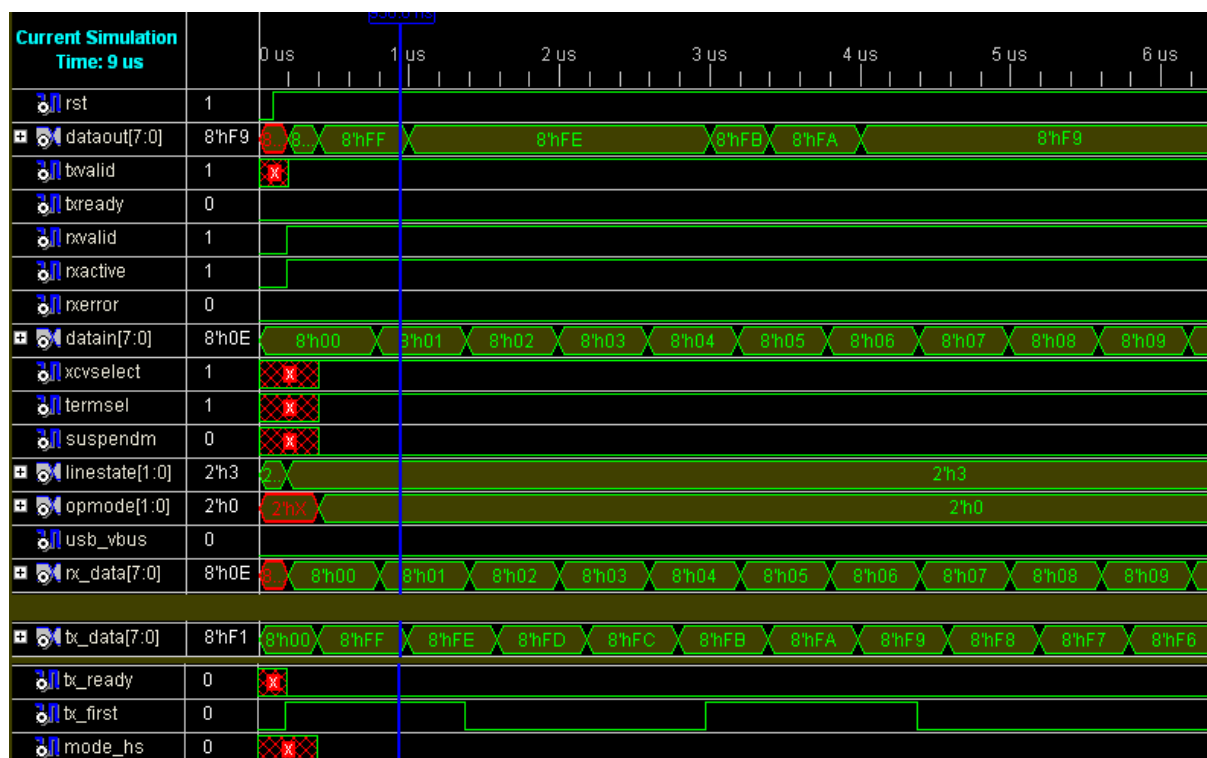


Figure 4.1 UTMI waveform

4.2 PACKET ASSEMBLY SIMULATION

The the waveform gives the packet simulation results. The input data was given using **tx_data** signal as Up-Counter. This signal comes from the buffer memory and then to be transmitted to UTMI. In the diagram, The CRC value of **8'h7C** gets attached to data. This CRC value is the lower byte of the signal **crc16_rev**. The lower byte got selected because **crc_sel2** is high.

After CRC, token PID gets attached responding to the control signal **send_token**. When that signal got asserted, token pid **8'h1E** got attached. And so on.

The CRC, data PIDs, token PIDs get attached depending on the status the control signals.



Figure 4.2 Packet Assembly waveform

4.3 PACKET DISASSEMBLY SIMULATION

Unlike Packet Assembly, in packet Disassembly, the reverse phenomenon occurs. The Data coming from the UTMI through the signal **rx_data**, is decoded to get the Packet IDs and Data IDs and then stored the buffer memory. In the waveform, the **rx_data** signal is considered as an Up-Counter hence it is not justified to expect for the PIDs to be in proper order. In the result we got the data PIDs and token PIDs matching with its corresponding PID values.

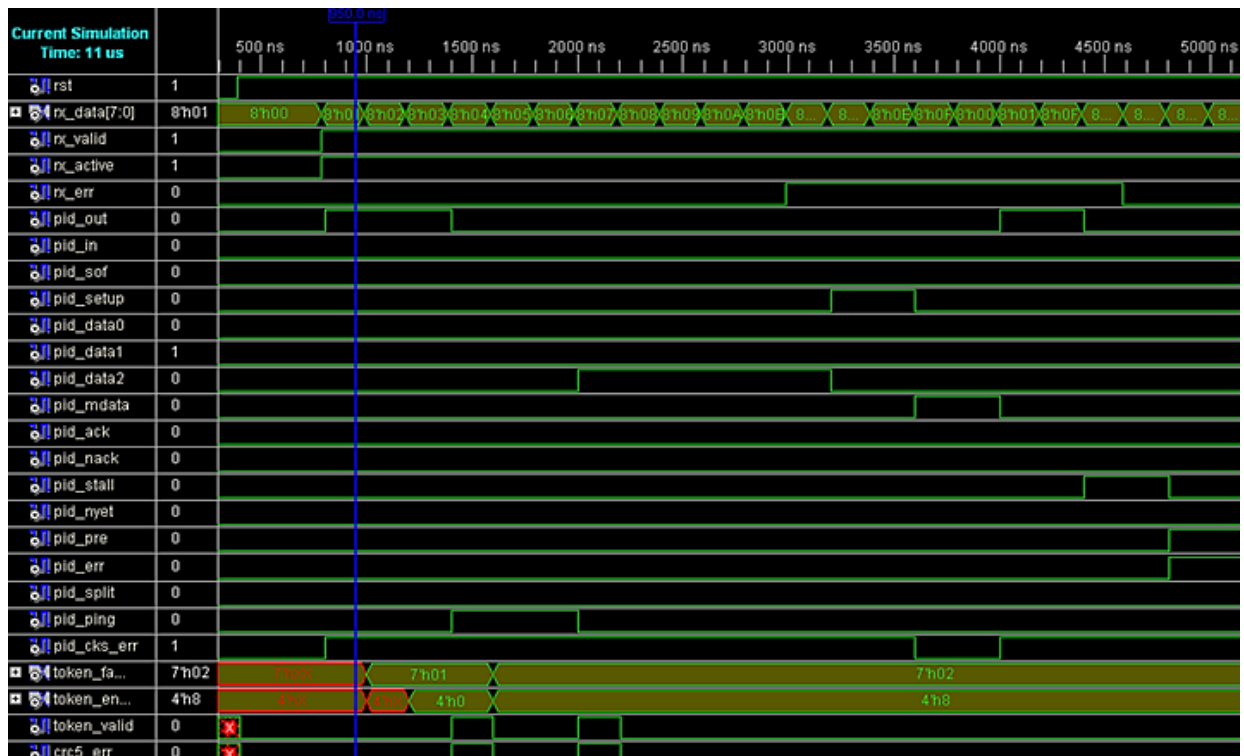


Figure 4.3 Packet Disassembly waveform

4.4 DATA BUFFERING SIMULATION

The data buffering simulation illustrates the storage method followed to fill up the buffer memory. The buffer memory is 32 bit in size, whereas the data coming in or out from UTMI is 8 bit in size. Hence it requires proper data organization. The input data considered were **rx_data** coming from the UTMI and **mdin** is the the internal memory. The **rx_data** is given as Up-Counter and **mdin** as random 32 bit data. 4 bytes of **rx_data** is collected and stored in **mdout**. Its value remains for the next 4 clocks and refilled with those 4 clocks data. Similarly the data in **mdin** gets split byte by byte for 4 clocks and moved to **tx_data**.

In the waveform **rx_data** values of 8'h04, 8'h05, 8'h06, 8'h07 gets stored in **mdout** as 32'h07060504. The **mdin** value 32'hA895C627 gets decoded in to 4 separate bytes and moved to **tx_data**. The location of the memory from which the data is to retrieved can be set using the **adr** signal.

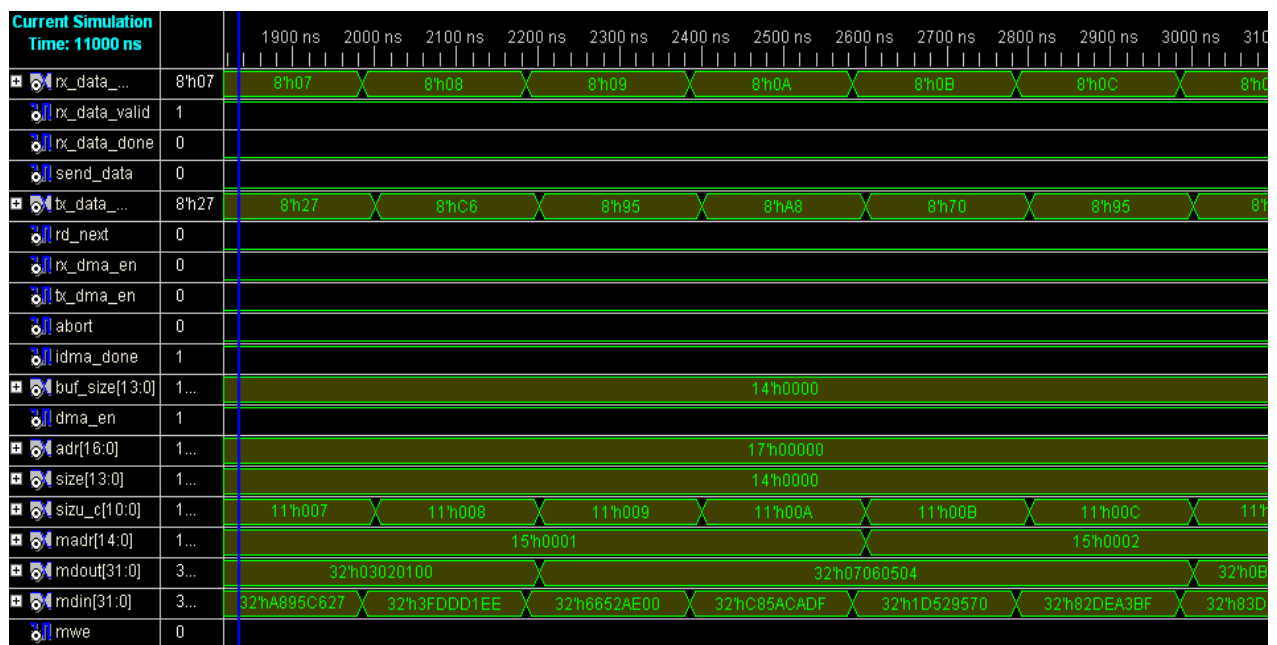


Figure 4.4 Data Buffering waveform

4.4 PROTOCOL ENGINE SIMULATION

For the simulation of Protocol Engine we have considered IN and OUT transaction in isochronous mode. These states were set using the CSR (Control Status Register) bits.

The following diagram shows the main flow chart.

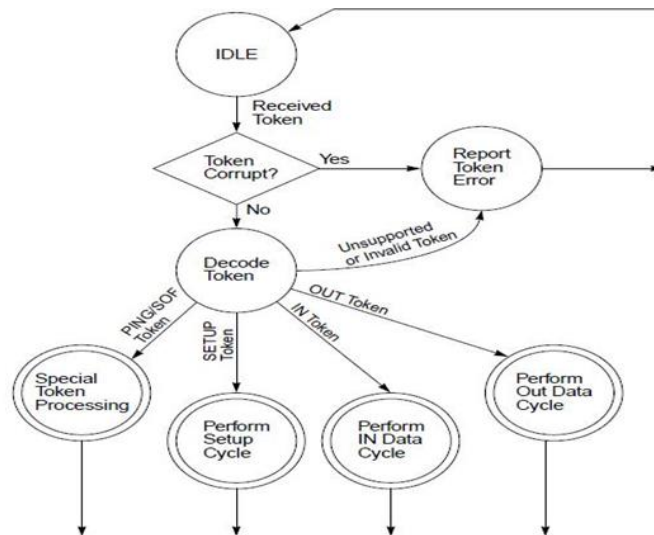


Figure 4.5 USB core main flow chart

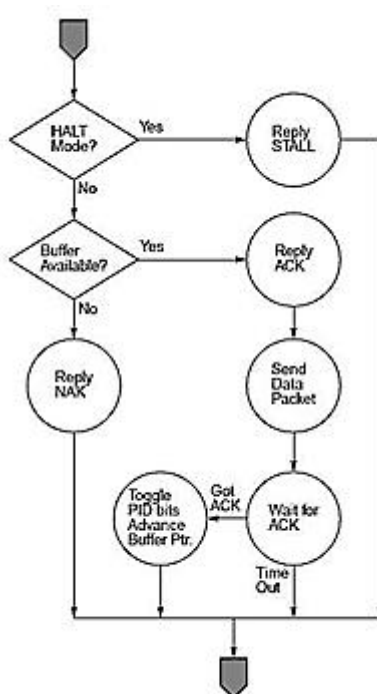


Figure 4.6 IN Cycle

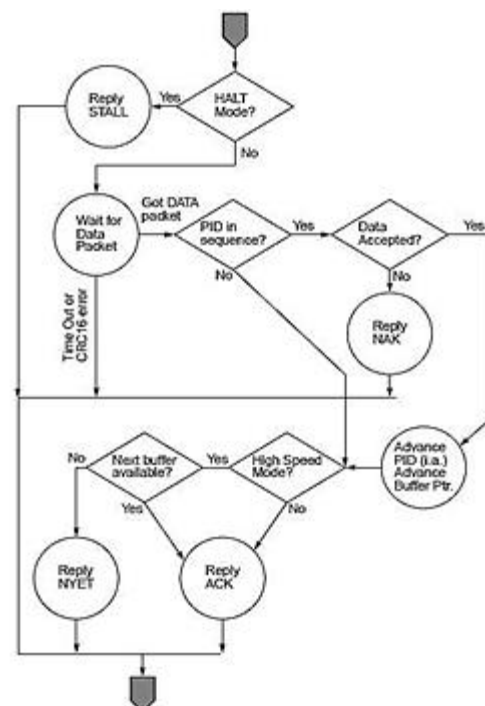


Figure 4.7 OUT cycle

In the protocol engine waveform initially an IN transaction was carried out. By this the **tx_dma_en** signal will get set. The buffer 1 gets activated and the address of buffer 1 (**h'88**) gets stored in **adr**. The size of buffer 1 gets stored in the signal **buf_size**. After getting into IN cycle, the data is sent. This is indicated by the **pid_data0** and hence change in **data_pid** signal was observed. After completion of 1 transaction we consider OUT transaction. The **rx_dmaen** signal get set. The buffet 0 gets activated and the address of buffer 0 (**h'80**) gets stored in **adr**. The size of buffer 0 gets stored in the **adr** and the process goes on. It was taken care that the MAX_PL_SZ (indicated by bit 10:0 of CSR) is less than the size of buffers. If the size becomes more than CSR then the transaction gets aborted. The signal **idin** and **state**, **next_state** signals updates the PID and transaction states

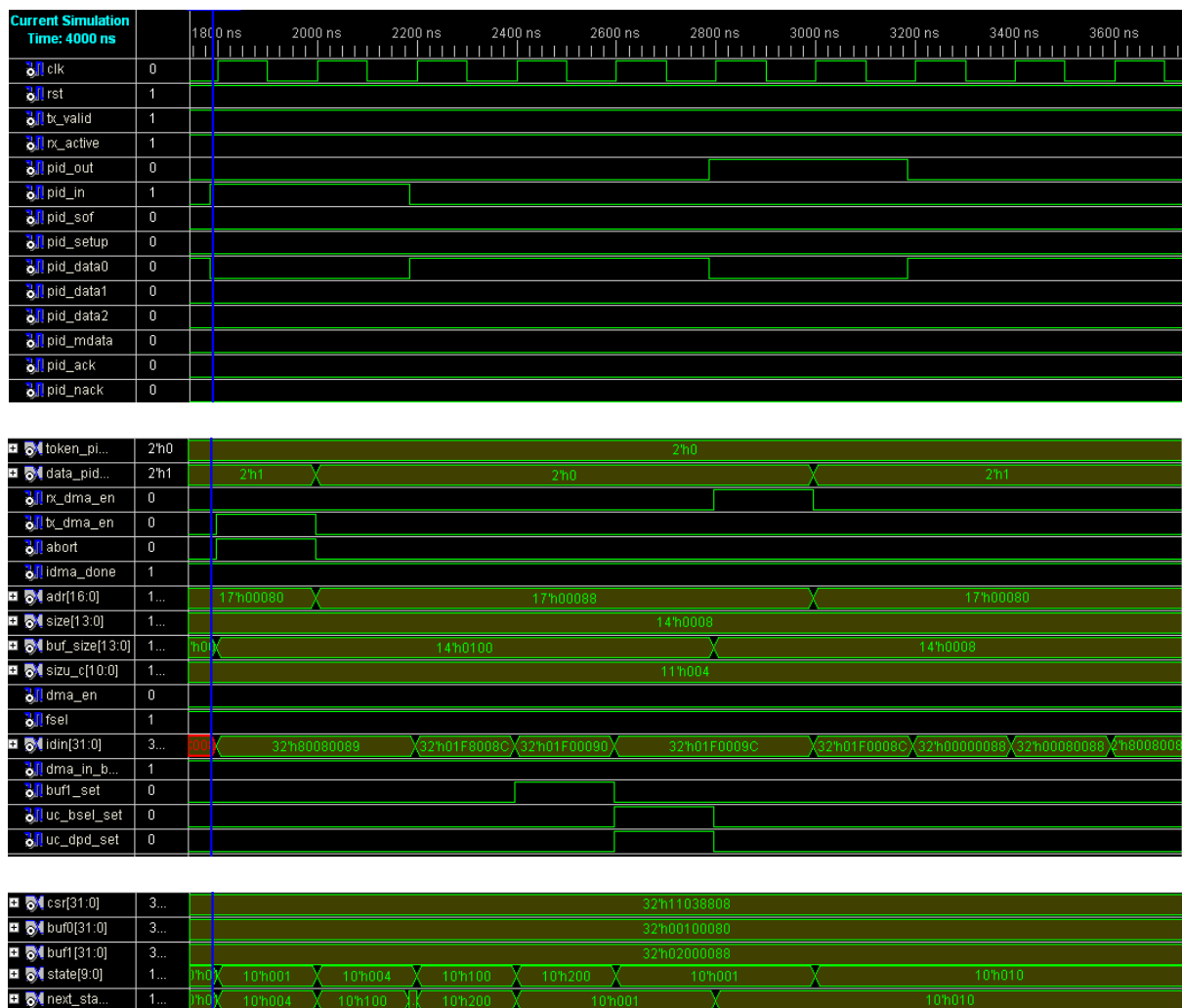


Figure 4.8 Protocol Engine waveform

Chapter 5

CONCLUSION

5.1 CONCLUSION

Universal Serial Bus (USB) was developed as a means to connect a large number of peripheral devices to the PC. USB 2.0 is an industry-wide, host oriented protocol, utilizing physical serial bus. It is host controlled i.e. there can be only 1 host per bus. The Communication flow complies with USB 2.0 Specifications essential for basic data transfer. The USB host is responsible for undertaking all transactions and scheduling the bandwidth. Data can be sent by various transaction methods using a token-based protocol. The USB Specification defines three data speeds: Low speed (1.5Mbit/s), Full speed (12 Mbit/s) and High speed (480Mbit/s).

With the introduction of USB 2.0 a new Host Controller Interface Specification, the EHCI (Enhanced Host Controller Interface) came into existence to keep device development costs down. It handles high speed transfers, and hand over low and full speed transfers to either OHCI or UHCI companion controllers. EHCI uses a NRZI (Non Return to Zero Invert) encoding scheme to send data with sync field to synchronize the host and receiver clocks.

We simulated a UTMI and protocol layer modules of the USB core in XILINX ISE 10.1. The results were verified with test bench generation. It is observed that UTMI is a serial bus interface which deals with USB protocols and signalling. The protocol engine also meets UTMI specification, generating control signals for UTMI transceiver interface according to the FSM states.

Chapter 6

FUTURE WORK

6.1 FUTURE WORK

To develop the whole USB core, other verilog modules for memory arbiter, register file, function interface needs to be designed, synthesized and simulated.

Since the USB 2.0 is widely used in many peripheral devices. It is well established and has already proved itself to be reliable and efficient. The protocol's strict restrictions assure that different devices are compatible in every machine. This assures that the Protocol Engine we designed can be used in various products that uses USB2.0 standard as their chosen protocol.

An advance version of the Protocol Engine would be an assimilation of the DMA Controller into it. Including the DMA Controller would simplify the interface with the device's bus and reduce chances of bus contention by upgrading the synchronization capabilities of the Protocol Engine in the USB device.

Furthermore, USB3.0 spec has been published and not long from now the computer industry will adopt this standard and move forward to the more advance protocol. A future improvement for the Protocol Engine is to support USB3.0 features and speed.

REFERENCES

- [1] Axelson, Jan, (2001), *USB Complete: Everything You Need to Develop Custom USB Peripherals*. Third Edition
- [2] Fatemeh Arbab Jolfaei et al., High Speed USB 2.0 Interface for FPGA Based Embedded Systems, Isfahan University of Technology, Department of Electrical & Computer Engineering
- [3] Universal Serial Bus Specifications Revision 2.0
- [4] www.usb.org [online]
- [5] "USB Made Simple", Available at : <http://www.usbmadesimple.co.uk/index.html>
- [6] "USB in a nutshell, making sense of the USB standard", Craig Peacock, Nov 2002, Available at: www.beyondlogic.org
- [7] USB 2.0 Transceiver Macrocell Interface (UTMI): Retrieved from http://www.intel.com/technology/usb/download/2_0_xcvr_macrocell_1_05.pdf
- [8] Jan Axelson, USB Complete - Everything you need to develop custom USB peripherals, Lakeview research, 3rd Edition, Aug 2005
- [9] "libUSB Manual", libUSB Project, Available at: <http://libusb.sourceforge.net/>
- [10] "Spartan-3 FPGA Family: Complete Data Sheet", Xilinx Corp., Aug 2005