

# PID CONTROLLER DESIGN FOR VARIOUS PLANT MODEL

A THESIS SUBMITTED IN PARTIAL FULFILLMENT OF

THE REQUIREMENTS FOR THE DEGREE OF

**Bachelor of Technology in**

**Electronics and Instrumentation Engineering**



By

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**Department of Electronics & Communication Engineering**

**National Institute of Technology, Rourkela**

**2014**

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Under the Guidance of

***Prof. T K DAN***



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**National Institute of Technology**

**Rourkela**



**National Institute of Technology  
Rourkela  
CERTIFICATE**

This is to certify that the thesis titled “**PID CONTROLLER DESIGN FOR VARIOUS PLANT MODEL**”, submitted by Mr RAJESH KUMAR (107EI008) in partial fulfilment of the requirements for the award of Bachelor of Technology Degree in ‘ELECTRONICS & INSTRUMENTATION’ Engineering during session of 2010-2014 at the National Institute of Technology (NIT), Rourkela is an authentic work carried out by him under my supervision.

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Rajesh kumar (110EI0504)

# CONTENTS

## ABSTRACT

- 1. INTRODUCTION**
- 2. PROCESS MODELING FROM RESPONSE CHARACTERISTICS OF PLANT**
  - 2.1 FOPDT**
    - 2.1.1 FINDING PARAMETERS OF FOPDT**
    - 2.1.2 STEP RESPONSE OF PLANT**
  - 2.2 IPDT**
  - 2.3 FOIPDT**
- 3. DIFFERENT TUNING PROCEDURE**
  - 3.1 TUNING FORMULA FOR FOPDT**
    - 3.1.1 Ziegler-Nichols tuning formula**
    - 3.1.2 Chine-Hrones-Reswick PID tuning algorithm**
    - 3.1.3 Cohen-Coon Tuning algorithm**
    - 3.1.4 Wang-Juang-Chan tuning formula**
    - 3.1.5 Optimal PID Controller Design**
  - 3.2 TUNING FORMULA FOR IPDT**
  - 3.3 TUNING FORMULA FOR FOIPDT**
- 4. SIMULATION OF FOPDT**
  - 4.1 SIMULATION OF Ziegler-Nichols tuning formula**
  - 4.2 SIMULATION OF Chine-Hrones-Reswick PID tuning algorithm**
  - 4.3 SIMULATION OF Cohen-Coon Tuning algorithm**
  - 4.4 SIMULATION OF Wang-Juang-Chan tuning formula**
  - 4.5 SIMULATION OF OPTIMUM CONTROLLER DESIGN**
- 5. SIMULATION OF IPDT**
- 6. SIMULATION OF FOIPDT**
- 7. CONCLUSION AND FUTURE WORKS**

## REFERENCE

# ABSTRACT

In this thesis I have discussed how to get a process model from response characteristics of a plant. Then I have discussed about various tuning formula used for finding controllers parameters. On the basis of the tuning formula P, PI, PD and PID controllers are designed and simulation is taken using Matlab and Simulink for different value of 'N' (filter coefficient) .

# Chapter 1

## INTRODUCTION

There are various types of industrial process. About process transfer function we know that finding a real value of it is very difficult. So whatever we get the transfer function of plants that can be approximately modelled by some definite transfer function. Some of those are FOPDT (first order plus delay time), IPDT (integral plus delay time) and FOIPDT (first order plus lag and integral delay time).

In thesis modeling of plant is done by using MATLAB. By Matlab we trace step response of plant. And using some basic calculation find out parameters of the model.

After finding model equation, next step is to find out controllers (P, PI, PD and PID) parameters. For this step we use various controller tuning method. For FOPDT model **Ziegler-Nichols tuning formula, Chine-Hrones-Reswick PID tuning algorithm, Cohen-Coon Tuning algorithm, Wang-Juang-Chan tuning formula** and **optimal PID controller design** are used for controller tuning. But controller used for IPDT and FOIPDT can't be tuned using these tuning formula, so we use different tuning formula for these model. For IPDT and FOIPDT only PD and PID controller is used.

After finding controller parameter response is taken in Simulink. For different value of filter coefficient.



Lastly observation is taken. For different controller rise time and settling time of response is noticed.

## Chapter 2

# PROCESS MODELING FROM RESPONSE CHARACTERISTICS OF PLANT

FOPDT (first order plus dead time), IPDT (integral plus dead time) and FOIPDT (first order lag and integrator plus dead time) are some basic plant model. In real time process control system a large variety of plant can be approximately model by FOPDT.

Equation of these model are:

FOPDT:

$$G(s) = K * e^{-Ls} / (Ts + 1)$$

IPDT:

$$G(s) = K e^{-Ls} / s$$

FOIPDT MODELS

$$G(s) = K e^{-Ls} / s * (Ts + 1)$$

Where

K=gain;

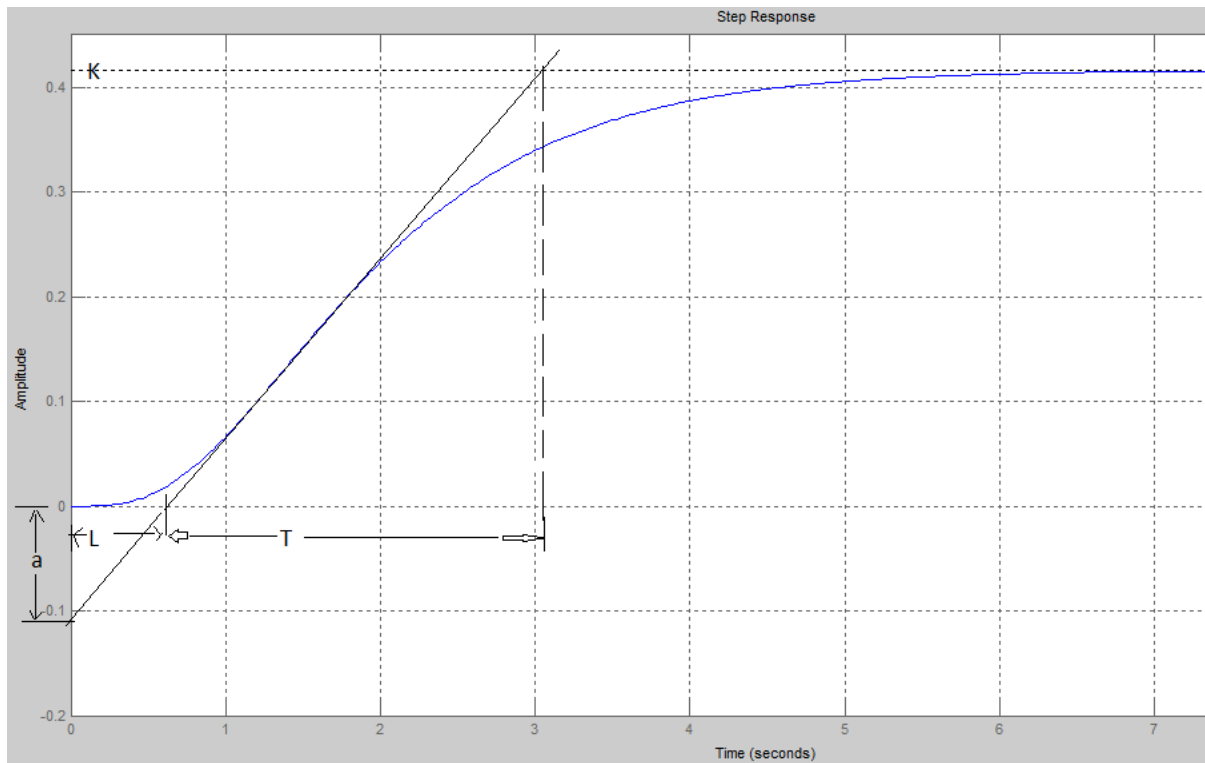
L= time delay;

T= time constant;

We can't derive the model physically and we need to perform the experiment to get the values of the parameters. Here Matlab is used to trace the response of plant versus time. For finding plant model parameter some basic calculation have to done.

## **2.1 FOPDT (first order plus dead time)**

For instance, if the step response of the plant model can be measured through an experiment, the output signal can be recorded as sketched below and from which the parameters of  $k$ ,  $L$ , and  $T$  (or  $a$ , where  $a = kL/T$ ) can be extracted by the simple approach shown.



### 2.1.1 Finding parameter of FOPDT

Let process transfer function of a plant is

$$G(s) = 10 / ((s+1)(s+2)(s+3)(s+4))$$

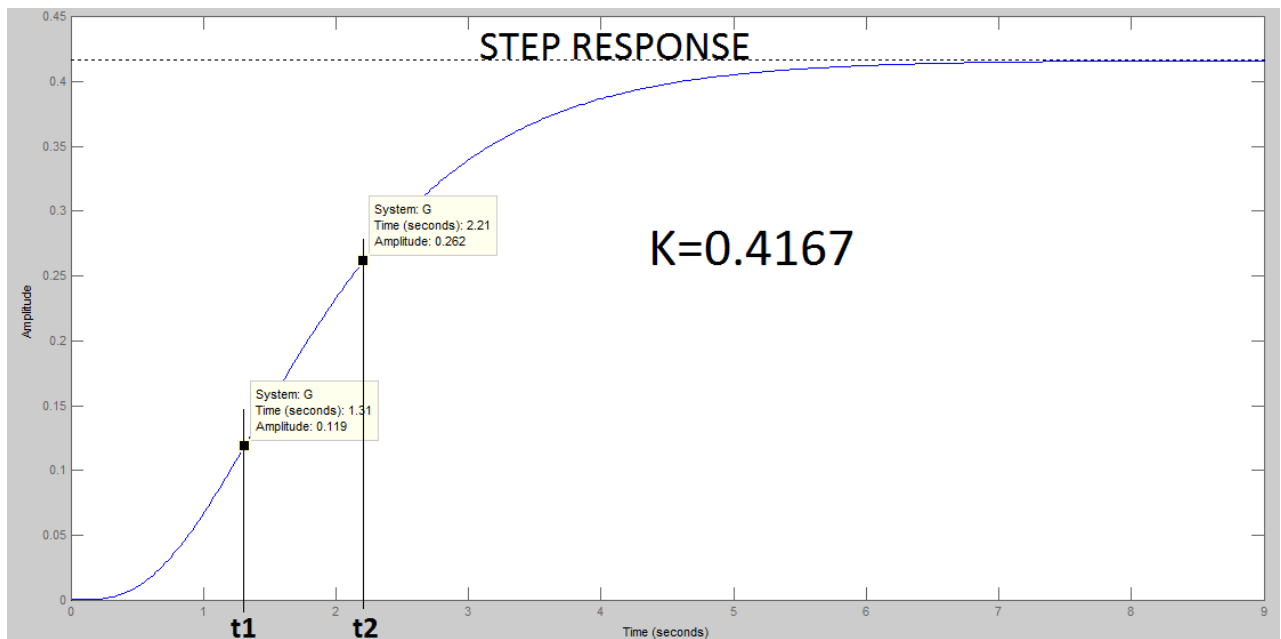
For gating step response of system a small matlab program is written:

```

1 -   clc;
2 -   close all;
3 -   clear all;
4 -   s=tf('s');
5 -   G=10/(s+1)/(s+2)/(s+3)/(s+4);
6 -   step(G);
7 -   k=dcgain(G)

```

### 2.1.2 Step Response of plant is



$t_1$ =time at which,  $gain(c) = 0.283 \cdot \text{steady state gain (K)}$

$t_2$ =time at which,  $gain(c) = 0.632 \cdot \text{steady state gain (K)}$

We have two equation for finding T and L

$$T=3(t_2-t_1)/2$$

$$L=(t_2-t_1)$$

$$a=KL/T$$

So from step response;

$$K=0.4167$$

$$t_1= 1.31 \text{ sec}$$

$$t_2=2.21 \text{ sec}$$

and

$$L=.855 \text{ sec}$$

$$T=1.365 \text{ sec}$$

Now we have FOPDT equation as:

$$G(s)= .4167 * e^{-.855s} / (1.365s + 1)$$

## 2.2 IPDT

Parameter of this plant model can be calculated by same procedure as FOPDT.

For this model let

$$K=.417$$

$$L=.855$$

So plant model equation is:

$$G(s) = 0.417 * e^{-.855s} / s$$

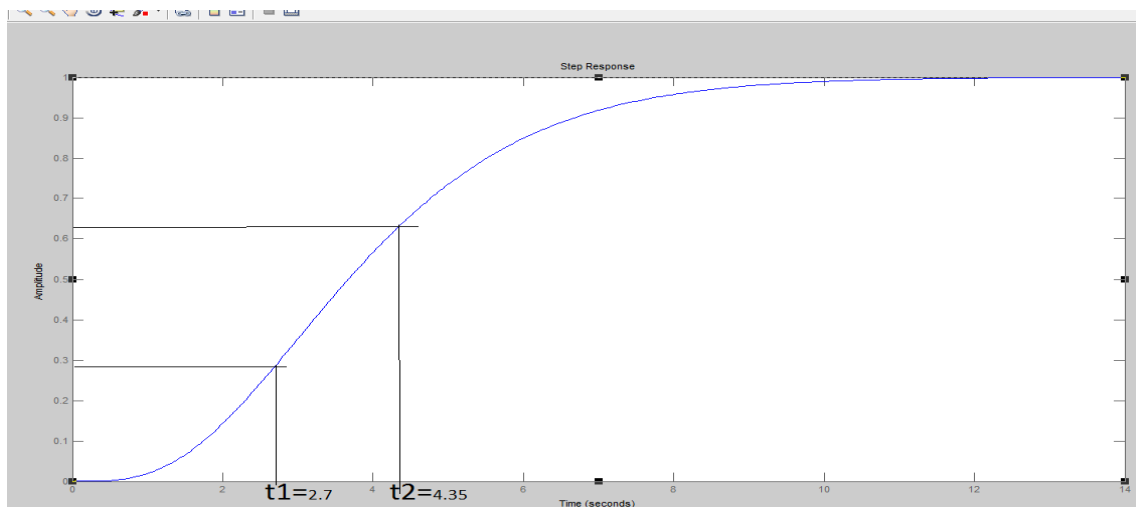
We don't need any extra integrator to remove the steady state error because there is already an integrator planted and error caused due to disturbances we use integrator for that process. Large overshoot can be ignored by PD controller.

## 2.3 FOIPDT

Parameters of FOIPDT is calculated using the procedure used in FOPDT model.

Let a process is  $G(s) = 1/s(s+1)^4$

Its step response is



So we may calculate  $k$ ,  $T$  &  $L$  with some simple calculation as above done

$$K=1$$

$$T=2.475$$

$$L=1.875$$

**Now** we have FOIPDT model equation as:

$$G(s) = Ke^{-1.875s}/s*(2.475s+1)$$

Since an integrator is contained in the model, an extra integrator is not necessary in the controller to remove the steady-state error due to set point change. So a PD controller may be used if there is no steady state disturbance at the plant. If steady state error due to disturbance is present then PID controller will be use



## Chapter 3

# **DIFFERENT TUNING PROCEDURE**

For finding controller parameters same tuning procedure can't be used for all types of plant model. For each plant model different tuning formula is used.

### **3.1 Tuning formula used for FOPDT**

>**Ziegler-Nichols tuning formula**

> **Chine-Hrones-Reswick PID tuning algorithm**

>**Cohen-Coon Tuning algorithm**

>**Wang-Juang-Chan tuning formula**

#### **3.1.1 Ziegler- Nichols tuning formula**

Ziegler and Nichols proposed this formula in 1942.

From the given formula one can find PID controller parameter either tracing step response of process transfer function or Nyquist plot of the transfer function.

From step response we find out 'L' and 'a' value and using this value in the formula we get PID parameter.

From Nyquist plot we get crossover frequency ( $\omega_c$ ) and the ultimate gain **K<sub>c</sub>** can be obtained. Then using these in formula controller parameter is found out.

The tuning formula is

Controller type	From step response			From frequency response		
	<u>Kp</u>	Ti	Td	<u>Kp</u>	Ti	Td
P	.3/a			.7/a		
PI	.6/a	4L		.7/a	2.3L	
PID	.95/a	2.4L	.42L	1.2/a	2L	.42L

Here only using step response; controller parameters are found out. Then using Simulink output step response the model plant is taken.

We have FOPDT equation as:

$$G(s) = .4167 * e^{-.855s} / (1.365s + 1)$$

So

$$a = KL/T = .19121$$

Controller....

$$P = 5.229$$

$$PI = 4.7069(1 + s/2.565)$$

$$PID = 6.2758 + 3.67006/s + 2.6829N/(1 + N/s)$$

### 3.1.2 Chien-Hrones-Reswick PID tuning algorithm

The Chien-Hrones-Reswick (CHR) method focus on the set-point regulation or disturbance rejection. Also regarding response speed and overshoot an additional

comment can be that Compared with the traditional Ziegler–Nichols tuning formula, the CHR method uses the time constant  $T$  of the plant explicitly.

The more heavily damped closed-loop response, which ensures, for the ideal plant model, the “quickest response without overshoot” is labeled “with 0% overshoot,” and the “quickest response with 20% overshoot” is labeled “with 20% overshoot”.

The CHR PID controller tuning formulas are given here for set point tracking and disturbance rejection.

### CHR tuning formulae

#### For set point regulation

Controller type	With 0% overshoot			With 20% overshoot		
	<u>Kp</u>	Ti	Td	<u>Kp</u>	Ti	Td
P	.3/a			.7/a		
PI	.35/a	1.2T		.6/a	T	
PID	.6/a	T	.5L	.95/a	1.4T	.47L

For disturbance rejection

Controller type	With 0% overshoot			With 20% overshoot		
	<u>Kp</u>	Ti	Td	<u>Kp</u>	Ti	Td
P	.3/a			.7/a		
PI	.6/a	4L		.7/a	2.3L	
PID	.95/a	2.4L	.42L	1.2/a	2L	.42L

□ 1

Now on the basis of mentioned table I have found out the controller parameter.

**For set point regulation:**

**For Zero % overshoot**

P controller:  $K_p = 1.5687$

PI controller:  $K_p = 1.8305$        $T_i = 1.638$

PID controller:  $K_p = 3.1374$        $T_i = 1.368$        $T_d = .4275$

**For 20 % overshoot**

P controller:  $K_p = 3.66$

PI controller:  $K_p = 3.137$        $T_i = 1.365$

PID controller:  $K_p = 4.9675$        $T_i = 1.91$        $T_d = .402$

**For disturbance rejection:**

**For Zero % overshoot**

P controller:  $K_p = 1.5687$

PI controller:  $K_p = 3.137$        $T_i = 3.42$

PID controller:  $K_p = 4.9675$        $T_i = 2.052$        $T_d = .3591$

**For 20 % overshoot**

P controller:  $K_p = 3.6603$

PI controller:  $K_p = 3.6603$        $T_i = 1.9665$

PID controller:  $K_p = 6.2748$        $T_i = 1.71$        $T_d = .3591$

### 1.1.3 Cohen-Coon Tuning algorithm

It is another type of Ziegler-nichols tuning formula.

The different controllers can be designed by the direct use of below table.

In this table 'a' and  $\tau$  experimentally calculated.

$$a = KL/T;$$

$$\tau = L/(L+T);$$

Controller	$K_p$	$T_i$	$T_d$
P	$\frac{1}{a} \left(1 + \frac{.35\tau}{1-\tau}\right)$		
PI	$\frac{.9}{a} \left(1 + \frac{.92\tau}{1-\tau}\right)$	$\frac{3.3-3\tau}{1+1.2\tau} L$	
PD	$\frac{1.24}{a} \left(1 + \frac{.13\tau}{1-\tau}\right)$		$\frac{.27-.36\tau}{1-.87\tau} L$
PID	$\frac{1.35}{a} \left(1 + \frac{.18\tau}{1-\tau}\right)$	$\frac{2.5-2\tau}{1-.39\tau} L$	$\frac{.37-.37\tau}{1-.81\tau} L$

### Controller parameters

P controller:  $K_p = 6.37536$

PI controller:  $K_p = 7.418$ ,  $T_i = 1.254$

PD controller:  $K_p = 7.019$ ,  $T_d = .1688$

PID controller:  $K_p = 7.855$ ,  $T_i = 1.74$ ,  $T_d = 0.2827$

### 1.1.4 Wang-Juang-Chan tuning formula

This tuning algorithm is proposed by Wang, Juang, and Chan. For selecting the PID parameters it is a simple and efficient method which is based on the optimum ITAE criterion. The controller parameters can given by, if the parameters K, L, T of the plant model are known,

$$K_p = (.7303 + .5307T/L)(T + .5L)/K(T + L)$$

$$T_i = T + .5L;$$

$$T_d = .5LT/(T + .5L)$$

Parameters value are

- $K_P = 3.0568$
- $T_i = 1.7925$
- $T_d = .3255$

### 1.1.5 Optimal PID Controller Design

Optimum setting algorithms for a PID controller were proposed by Zhuang and Atherton for various criteria. Consider the general form of the optimum criterion

$$J_n(\theta) = \int_0^{\infty} [t^n E(\theta, t)]^2 dt$$

Where  $e(\theta, t)$  is the error signal which enters the PID controller, with  $\theta$  the PID controller parameters. Two setting strategies for PID controller are proposed:

One for the set-point input and the other for the disturbance signal  $d(t)$ . In particular, three values of  $n$  are discussed, i.e., for  $n = 0, 1, 2$ . These three cases correspond, respectively, to three different optimum criteria: the integral squared error (ISE) criterion, integral squared time weighted error (ISTE) criterion, and the integral squared time-squared weighted error (IST2E) criterion. The expressions given were obtained by fitting curves to the optimum theoretical results.

### **Set-Point optimum PID tuning**

#### **For PI controller**

$$K_p = (a_1/K)(L/T)^{b_1}$$

$$T_i = T / (a_2 + b_2(L/T))$$

#### **For PID controller**

$$K_p = (a_1/K)(L/T)^{b_1}$$

$$T_i = T / (a_2 + b_2(L/T))$$

$$T_d = a_3 T (L/T)^{b_3}$$



*Set point PI controller parameter*

Range of L/T	0.1 -1			1.1-2		
Criterion	ISE	ISTE	IST <sup>2</sup> E	ISE	ISTE	IST <sup>2</sup> E
a <sub>1</sub>	0.98	0.712	0.569	1.072	0.786	0.628
b <sub>1</sub>	-0.892	-0.921	-0.951	-0.560	-0.559	-0.583
a <sub>2</sub>	0.690	0.968	1.023	0.648	0.883	1.007
b <sub>2</sub>	-0.155	-0.247	-0.179	-0.114	-0.158	-0.167

*Set-point PID controller parameters*

Range of L/T	0.1-1			1.1-2		
Criterion	ISE	ISTE	IST <sup>2</sup> E	ISE	ISTE	IST <sup>2</sup> E
a <sub>1</sub>	1.048	1.042	0.968	1.154	1.142	1.061
b <sub>1</sub>	-0.897	-0.897	-0.904	-0.567	-0.579	-0.583
a <sub>2</sub>	1.195	0.987	0.977	1.047	0.919	0.892
b <sub>2</sub>	-0.368	-0.238	-0.253	-0.220	-0.172	-0.165
a <sub>3</sub>	0.489	0.385	0.316	0.490	0.384	0.315
b <sub>3</sub>	0.888	0.906	0.892	0.708	0.839	0.832

*Set-point PID controller parameters with D in feedback path*

Range of L/T	0.1-1			1.1-2		
Criterion	ISE	ISTE	IST <sup>2</sup> E	ISE	ISTE	IST <sup>2</sup> E
a <sub>1</sub>	1.260	1.053	0.942	1.295	1.120	1.001
b <sub>1</sub>	-0.887	-0.930	-0.933	-0.619	-0.625	-0.624
a <sub>2</sub>	0.701	0.736	0.770	0.661	0.720	0.754
b <sub>2</sub>	-0.147	-0.126	-0.130	-0.110	-0.114	-0.116
a <sub>3</sub>	0.375	0.349	0.308	0.378	0.350	0.308
b <sub>3</sub>	0.886	0.907	0.897	0.756	0.811	0.813

## Disturbance rejection PID controller

### PI controller

$$K_p = (a_1/T)(L/T)^{b_1}$$

$$T_i = (T/a_2)*(L/T)^{b_2}$$

### PID controller

$$K_p = (a_1/T)*(L/T)^{b_1}$$

$$T_i = (T/a_2)(L/T)^{b_2}$$

$$T_d = a_3 T(L/T)^{b_3}$$

*Disturbance rejection PI controller parameter*

Range of L/T	0.1 -1			1.1-2		
Criterion	ISE	ISTE	IST <sup>2</sup> E	ISE	ISTE	IST <sup>2</sup> E
a <sub>1</sub>	1.279	1.015	1.021	1.346	1.065	1.076
b <sub>1</sub>	-0.945	-0.957	-0.953	-0.675	-0.673	-0.648
a <sub>2</sub>	.535	.667	.629	.552	.687	.650
b <sub>2</sub>	.586	.552	.546	.438	.427	.442

*Disturbance rejection PID controller parameters*

Range of L/T	0.1-1			1.1-2		
Criterion	ISE	ISTE	IST <sup>2</sup> E	ISE	ISTE	IST <sup>2</sup> E
a <sub>1</sub>	1.473	1.468	1.531	1.524	1.515	1.592
b <sub>1</sub>	-0.97	-0.97	-0.96	-0.735	-0.73	-0.705
a <sub>2</sub>	1.115	.942	.971	1.13	.957	.957
b <sub>2</sub>	.753	.725	.746	.641	.598	.597
a <sub>3</sub>	.55	.443	.413	.552	.444	.414
b <sub>3</sub>	.948	.939	.933	.851	.847	.850

**Calculated controller parameter on the basis of this tuning:**

**For set point tracking:**

Pi controller

Criterion	<u>K<sub>p</sub></u>	Ti	Td
ISE	3.5696	2.3022	
ISTE	2.629	1.678	
ISTSE	2.1306	1.498	

PID controller

Criterion	<u>Kp</u>	Ti	Td
ISE	3.826	1.415	.4406
ISTE	3.8044	1.629	.31166
ISTSE	3.546	1.668	.3404

PID controller with D in feedback path

Criterion	<u>Kp</u>	Ti	Td
ISE	4.579	2.2417	.3382
ISTE	3.904	2.0744	.3117
ISTSE	3.498	1.982	.276

**For disturbance rejection**

Pi controller

Criterion	<u>Kp</u>	Ti	Td
ISE	1.458	1.9396	
ISTE	1.1635	1.5806	
ISTSE	1.1682	1.6809	

PID controller

Criterion	<u>Kp</u>	Ti	Td
ISE	1.693	.8607	.482
ISTE	1.693	1.0322	.3898
ISTSE	1.757	.992	.364

### 3.2 Tuning formula for IPDT

we have IPDT equation as:

$$G(s) = 0.417 * e^{-.855s} / s$$

As above discussed for IPDT we use PD and PID controller

#### The coefficient of the controller for IPDT models

Criterion	a1	a2	a3	a4	a5
ISE	1.03	.49	1.37	1.49	.59
ISTE	.96	.45	1.36	1.66	.53
ISTSE	.9	.45	1.34	1.83	.49

**Controller parameters calculated as:**

PD controller  $K_p = a1/KL$ ;  $T_d = aL$

PID controller  $K_p = a3/KL$ ;  $T_i = a4L$ ;  $T_d = a5L$

**Controller parameter:**

**PD controller**

Criterion	<u>Kp</u>	Ti	Td
ISE	2.89		.4189
ISTE	2.6944		.3847
ISTSE	2.526		.3847

**PID controller**

Criterion	<u>Kp</u>	Ti	Td
ISE	3.8451	1.274	.5044
ISTE	3.817	1.419	.453
ISTSE	3.7609	1.565	.4189

### 3.3 Tuning formula for FOIPDT

we have FOIPDT equation as:

$$G(s) = Ke^{-1.875s}/s*(2.475s+1)$$

A PD controller setting algorithm is

$$Kp = 2/3KL$$

$$Td = T$$

And PID setting algorithm

$$Kp=1.111T/(KL^2(1+(T/L)^{.65})^2); \quad Ti=2L(1+(T/L)^{.65})$$

$$Td=Ti/4$$

**Calculated Controller parameters are:**

**PD CONTROLLER:** Kp= .356,                      Kd=2.475

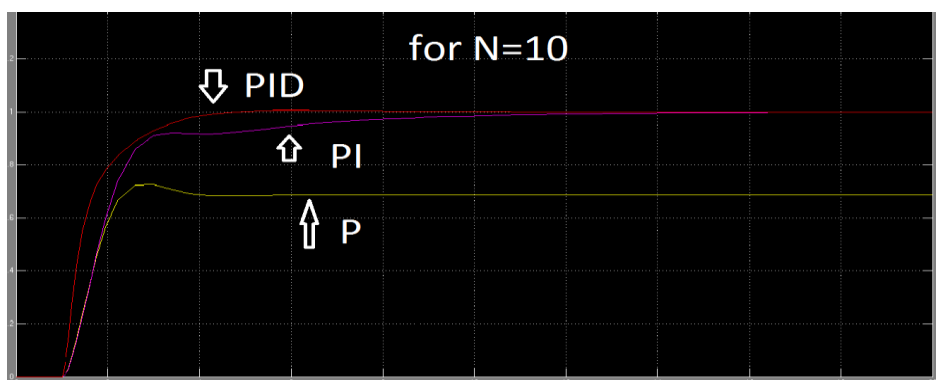
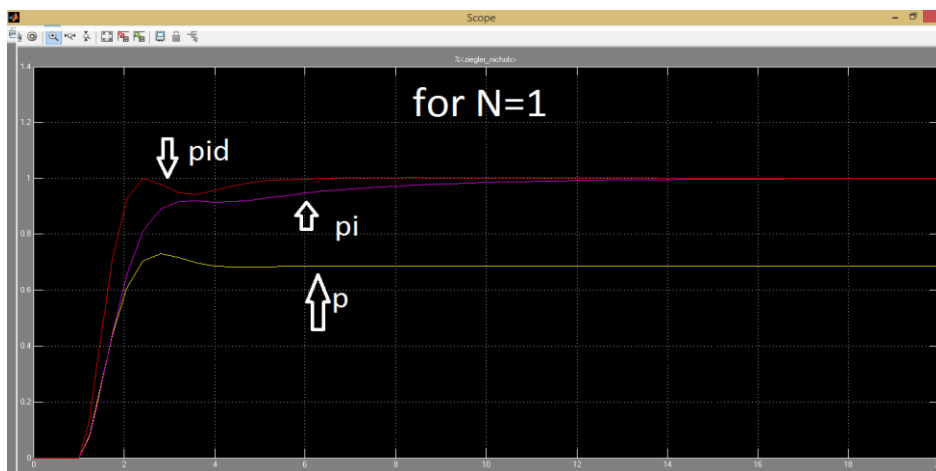
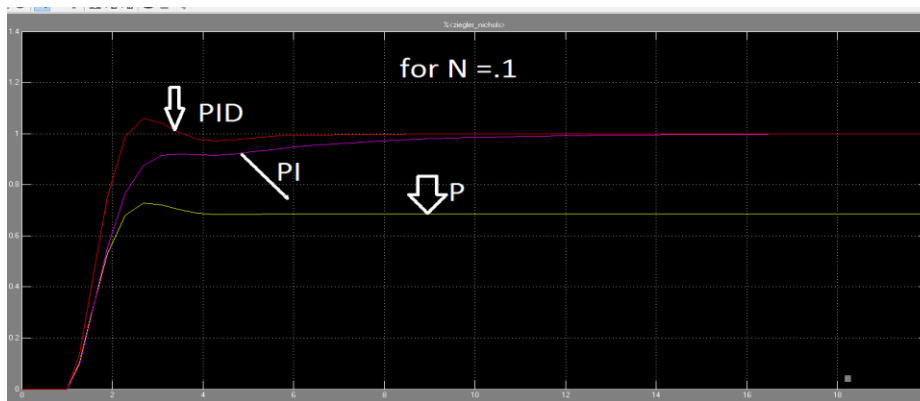
**PID CONTROLLER:** Kp=1617,      Ti=8.2416,      Td=2.04

# Chapter 4

## Simulation of FOPDT

Simulation is done using Simulink. Using tuning formula we have found out P, PI, PD and PID controller parameters. Response for FOIPDT plant modal is observed for different value of filter coefficient 'N'.

#### 4.1 simulation for Ziegler-Nichols tuning formula



Note

y axis : amplitude

x axis : time(s)

**Observation:**

**For N=10**

<b>Controller type</b>	<b>Rise time</b>	<b>Settling time</b>
<b>P</b>	<b>1.5</b>	<b>3</b>
<b>PI</b>	<b>2</b>	<b>11</b>
<b>PID</b>	<b>2.5</b>	<b>3.5</b>

**Result:**

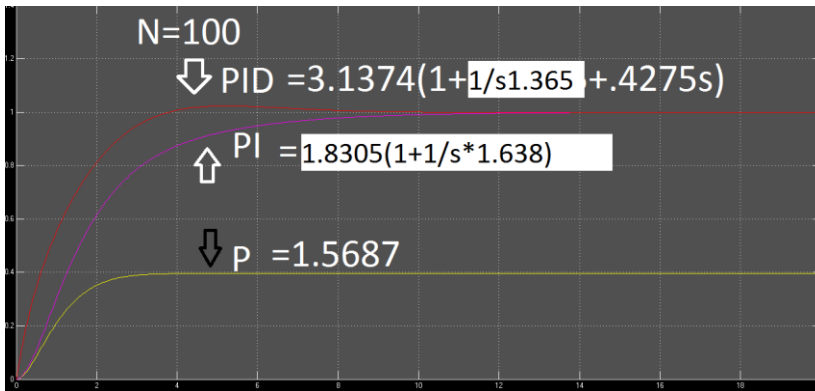
- So one can see that small value of 'N' increase overshoot, settling time and oscillation but reduce the rise time for PID controller.
- PI controller is eliminating the error but P controller is giving offset.

## **4.2 Simulation for Chine –Hrones-Reswick PID tuning algorithm**

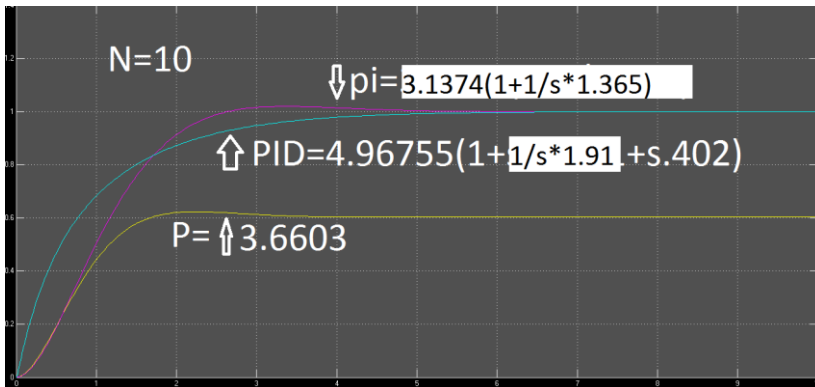
N=100



## Set point regulation for Zero overshoot

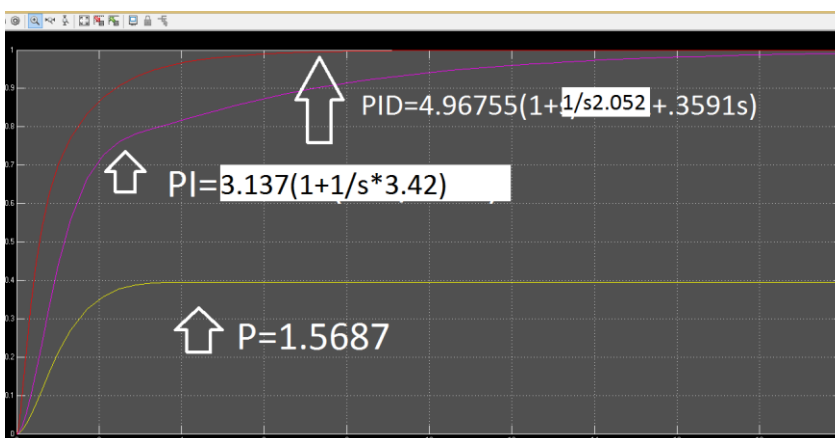


## For 20% overshoot

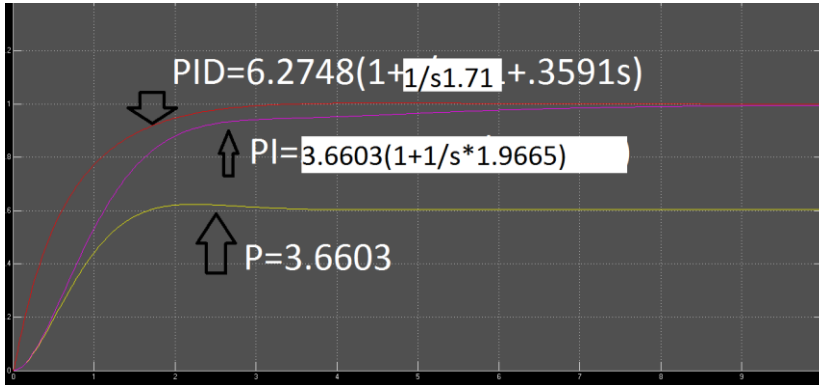


## Disturbance rejection for

### zero overshoot



20% overshoot



**Note:**

Y axis: amplitude

X axis: time

Observation:

For set point regulation:

Zero % overshoot

Controller type	R.T	S.T
P	2	3
PI	3.5	9
PID	2.5	8

20% overshoot

Controller type	R.T	S.T
P	1.5	3.5
PI	1.7	5
PID	2.4	6

For disturbance rejection

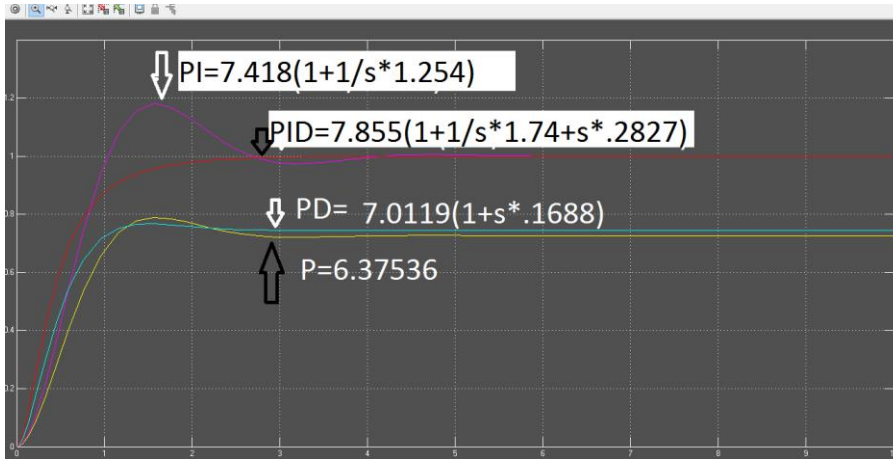
Zero% overshoot

Controller type	R.T	S.T
P	2	3
PI	7.75	20
PID	2	7

20% overshoot

Controller type	R.T	S.T
P	1.5	3
PI	2	8
PID	1.8	3

### 4.3 Simulation of Cohen-Coon Tuning algorithm



Note:

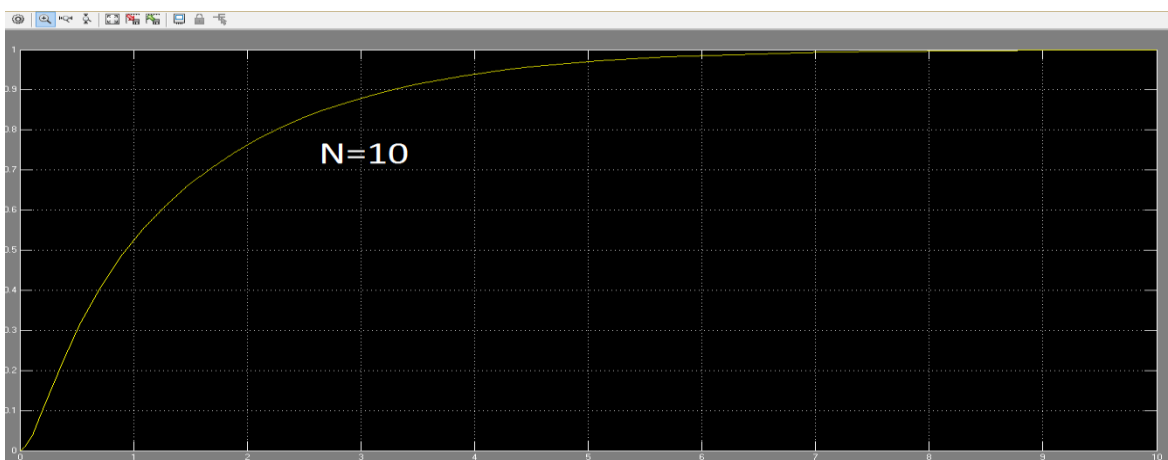
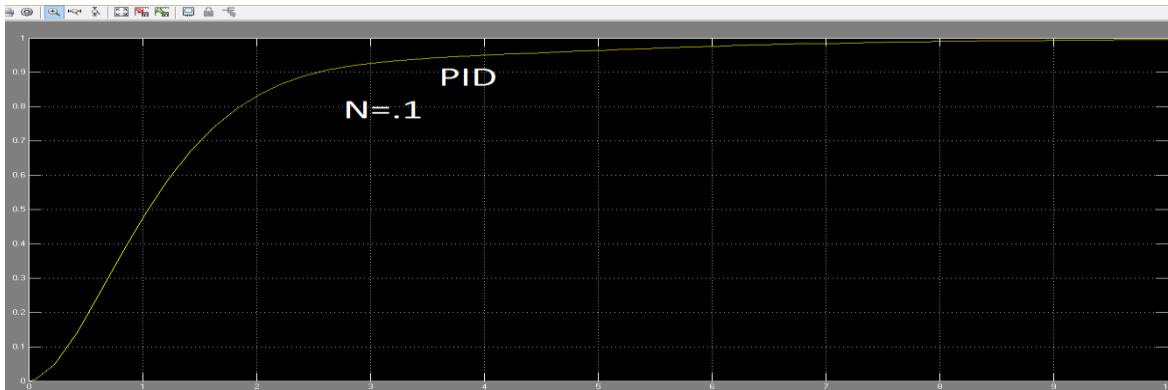
Y axis : amplitude

X axis: time(s)

Observation:

Controller type	R.T	S.T
P	.9	2.75
PI	1.2	4
PD	.7	2
PID	1	2.5

## 4.4 Simulation of Wang-Juang-Chan tuning formula



Note: Y axis: amplitude

X axis: time(s)

Observation: FOR N= 10;

Controller type	R.T	S.T
PID	3	7

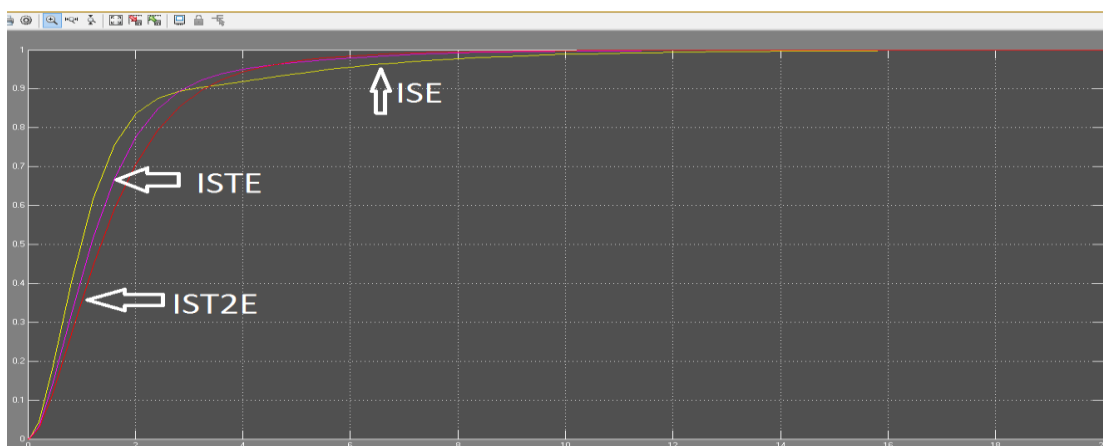
Result:

- small value of  $N$  reduced the rise time but increased the settling time.

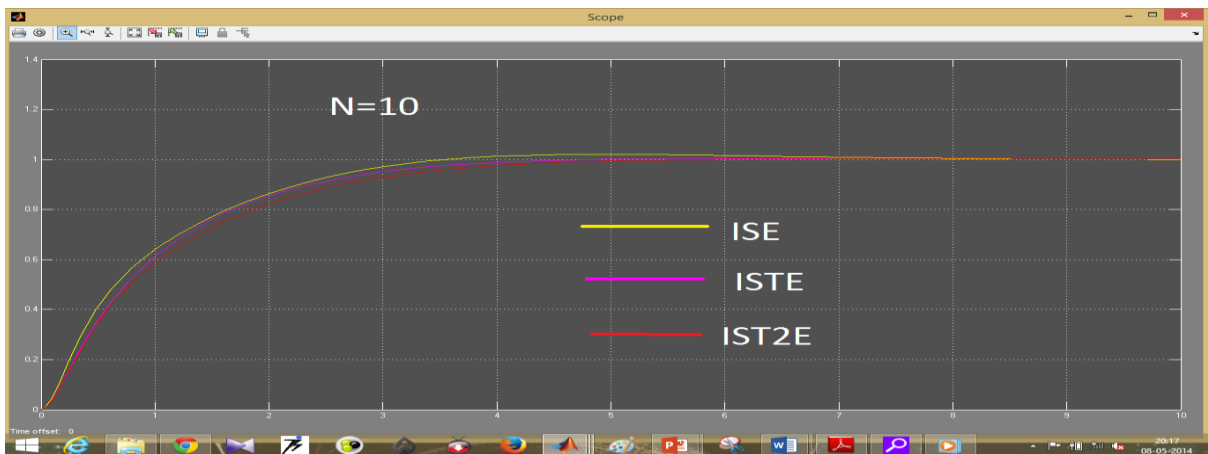
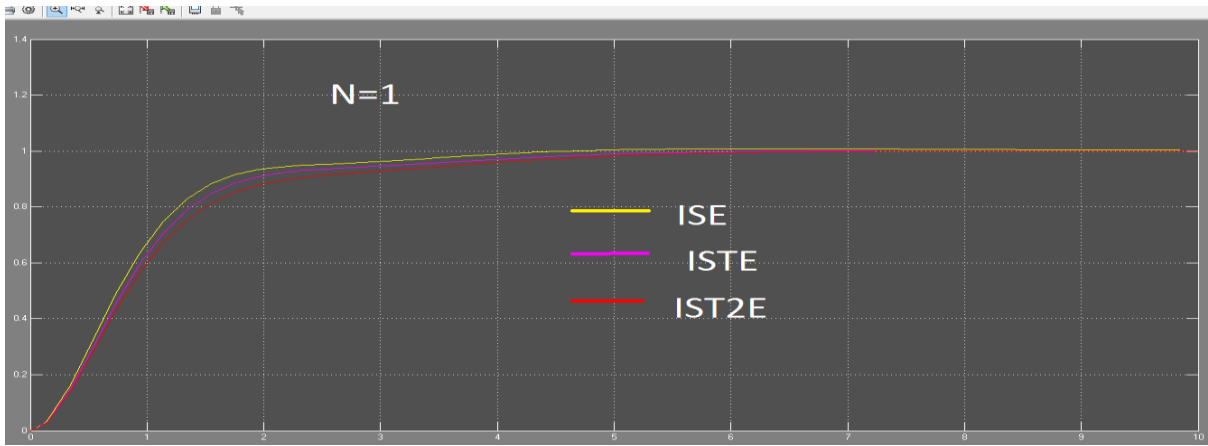
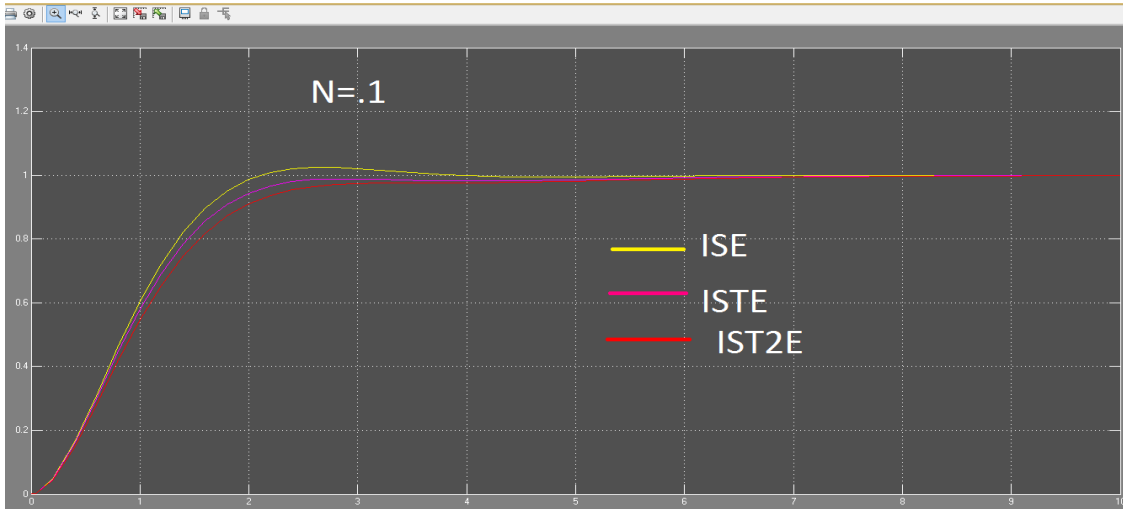
## 4.5 Simulation of Optimal PID Controller Design

For Set point tracking:

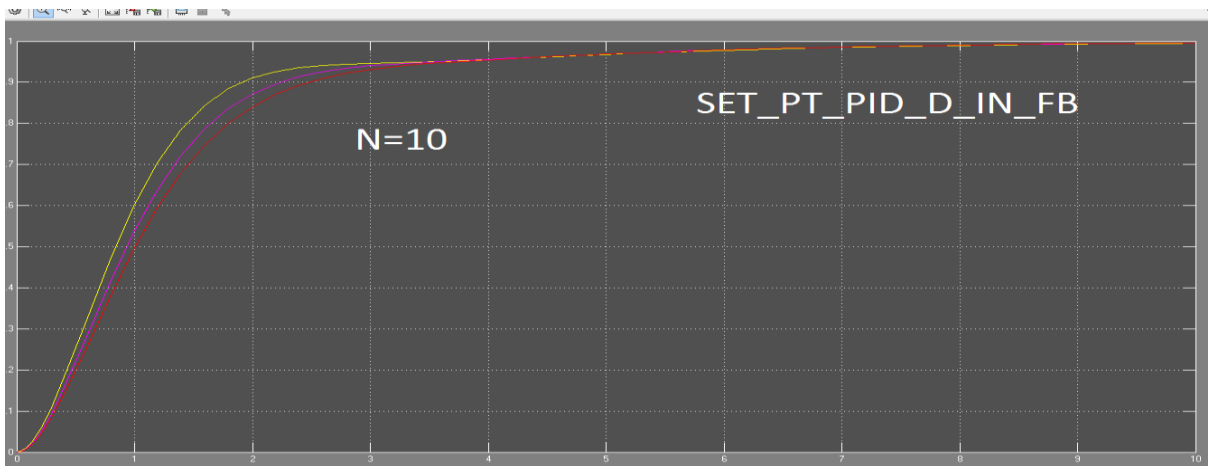
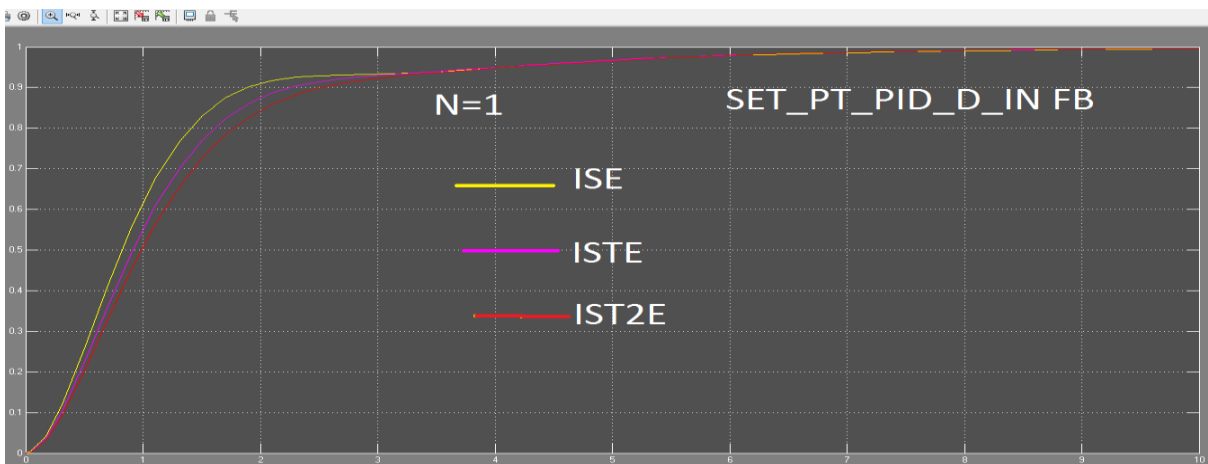
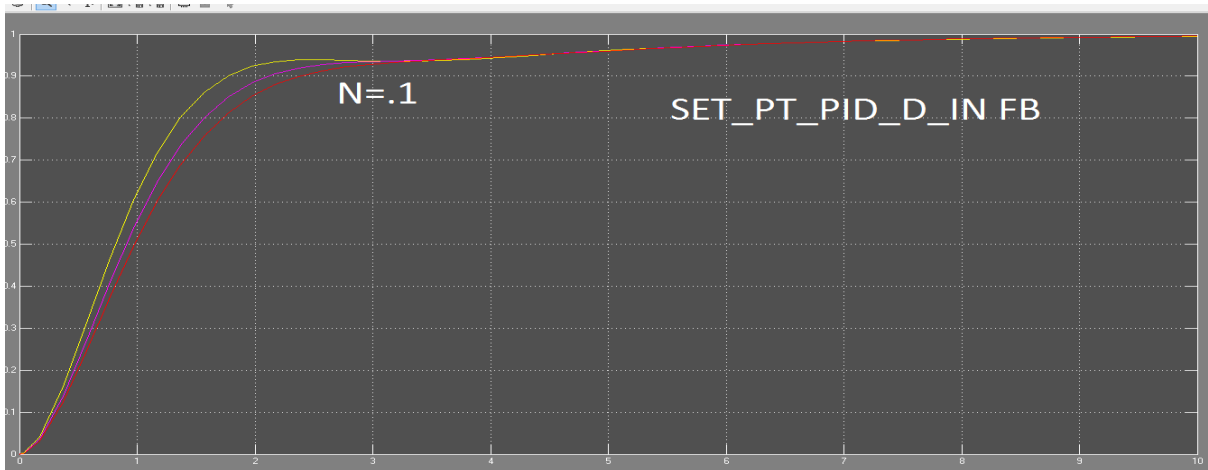
PI controller



PID controller



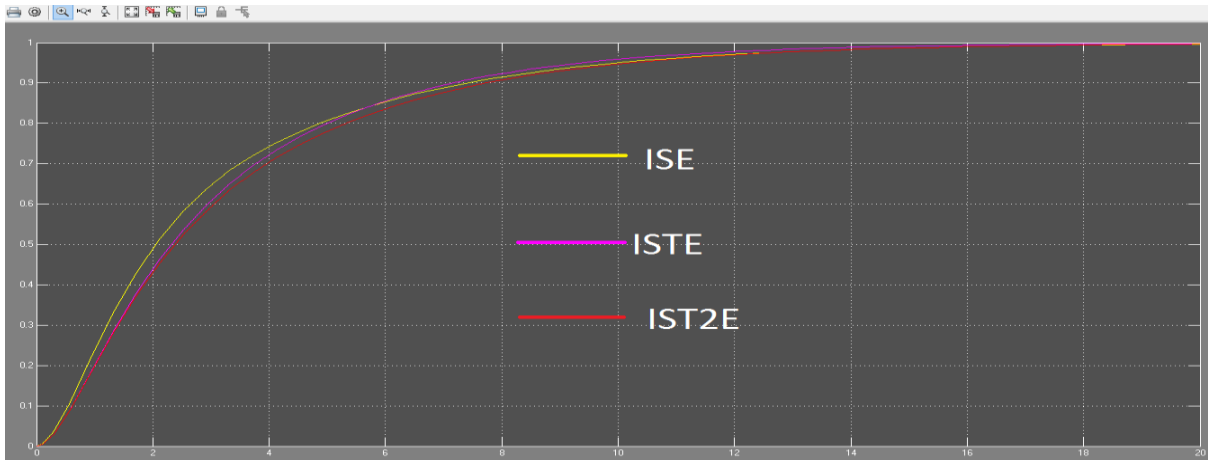
PID controller with D in feedback



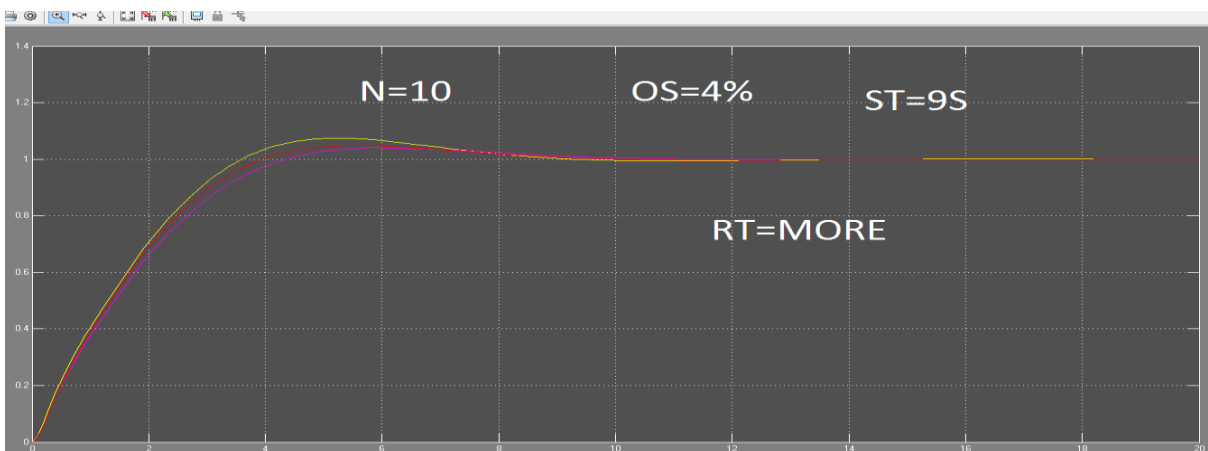
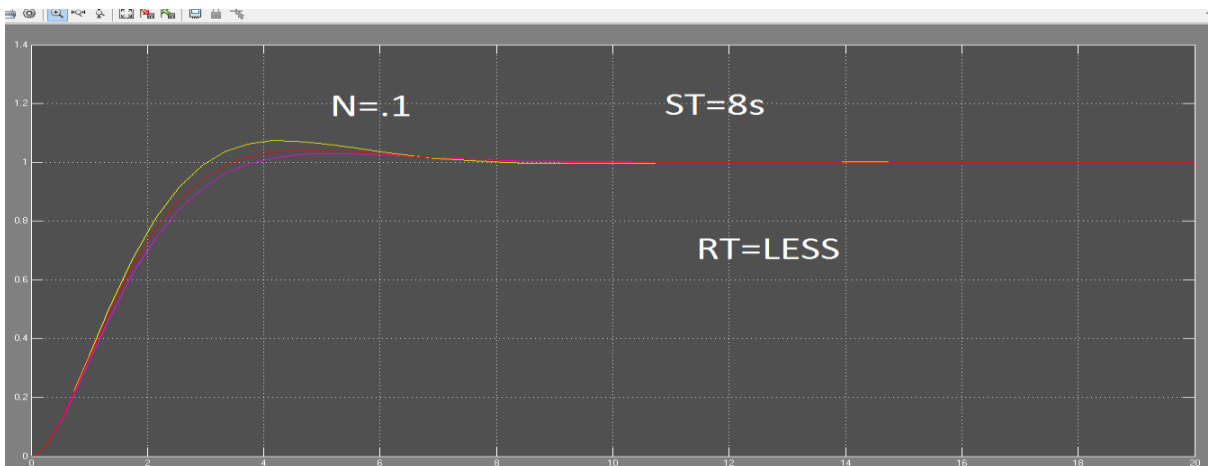
**Disturbance rejection:**



## PI controller



## PID controller



Observation:

For set point tracking:

PI controller:

Criterion type	R.T	S.T
ISE	2.8	12
ISTE	2.6	10
IST2E	3	8

PID controller: FOR N = 10

Criterion type	R.T	S.T
ISE	2.1	6.5
ISTE	2.3	4.5
IST2E	3	5

PID controller with D in feedback:

Criterion type	R.T	S.T
ISE	1.6	10
ISTE	2	10
IST2E	2.2	10

For disturbance rejection:

PI controller:

Criterion type	R.T	S.T
ISE	7	15.8
ISTE	6.4	16
IST2E	7.1	16.2

PID controller: N=10

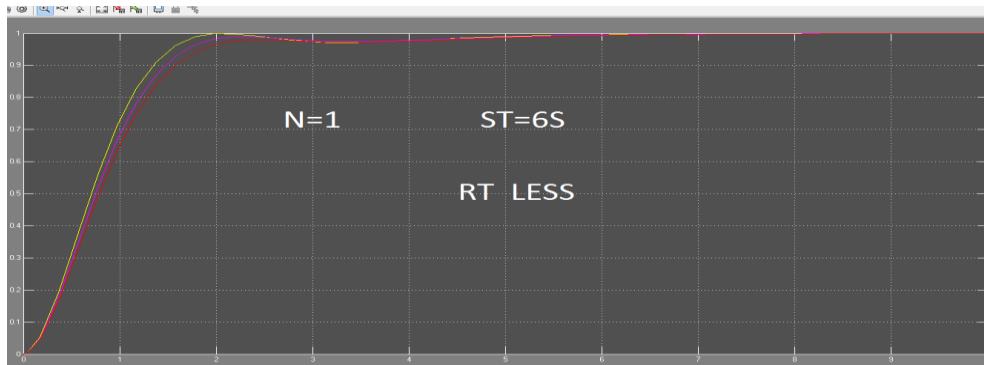
Criterion type	R.T	S.T
ISE	3	7
ISTE	3.4	7.2
IST2E	3.2	7.4

# Chapter 5

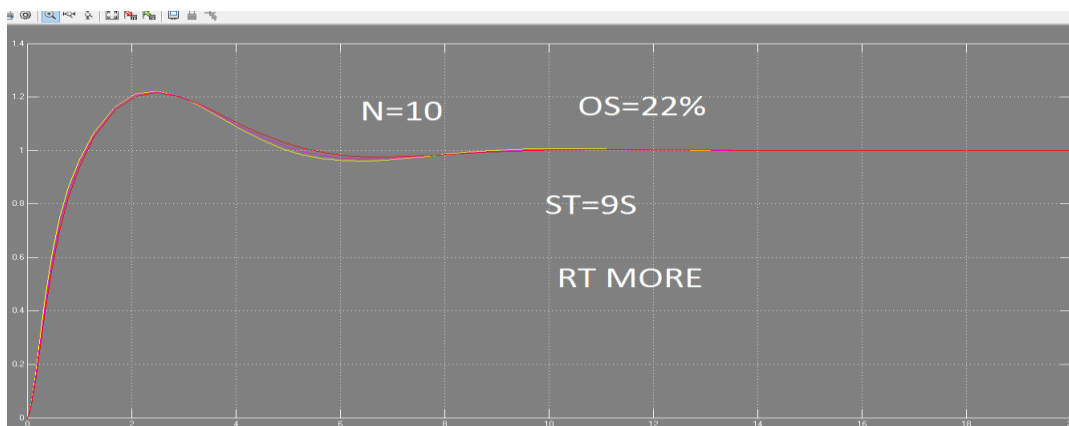
## Simulation of IPDT model

As above discussed for IPDT we use PD and PID controllers. Simulation taken corresponding these controller are given below.

### PD controller



### PID controller



**Note:**

**Y axis: amplitude**

**X axis :time()**

**Observation:**

**For PD controller**

Criterion type	R.T	S.T
ISE	1.8	4.9
ISTE	1.9	5
IST2E	2	5.1

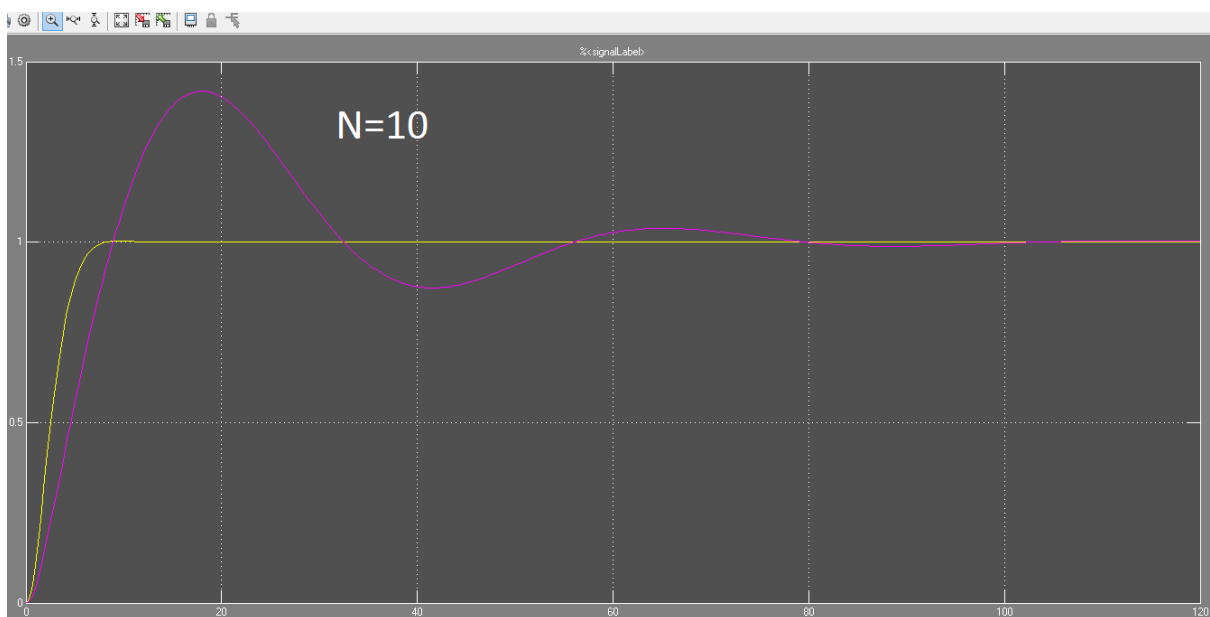
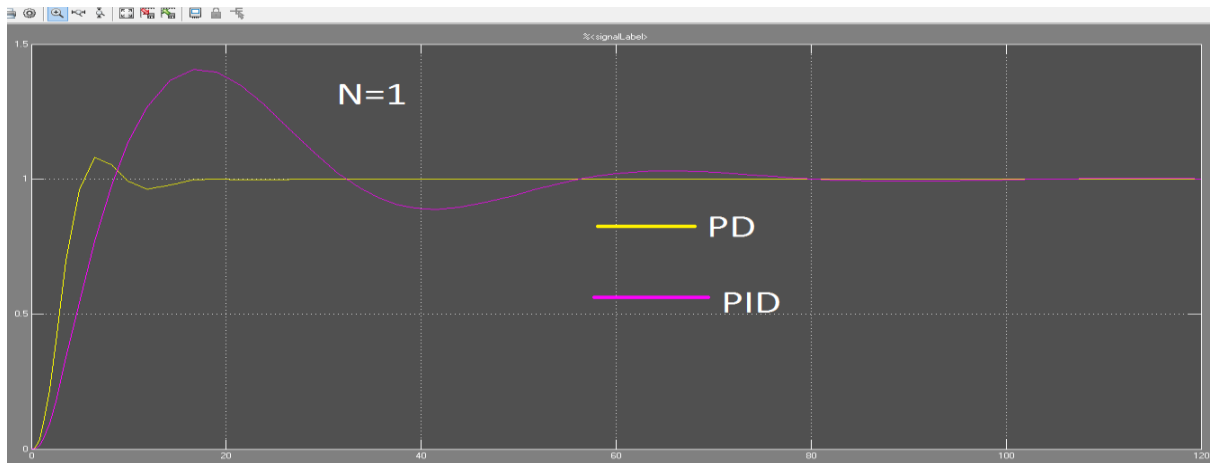
**For PID controller:**

Criterion type	R.T	S.T
ISE	1.6	8.8
ISTE	1.6	9
IST2E	1.6	9.2

## Chapter 6

# Simulation of FOIPDT MODELS

For FOIPDT PD and PID type of controller are used. Simulation regarding these controller shown below for different value of N.



**Note:**

**Y axis: amplitude**

**X axis: time(s)**



Observation: for N=10

Controller type	R.T	S.T
PD	4	8
PID	12	100

# **Chapter 7**

## **Conclusion and**

## **Future work**

Project study on PID controller design for various plant model provide a brief idea of plant modeling, type of plant model and controllers (P, PI, PD and PID) tuning method used for the of the model plant.

Plant model FOPDT, IPDT and FOIPDT are discussed in details. For each model different tuning methods are used. They help in finding controllers parameters. For tuning of controllers of FOPDT **Ziegler-Nichols tuning formula, Chinese-Hrones-Reswick PID tuning algorithm, Cohen-Coon Tuning algorithm, Wang-Juang-Chan tuning formula** and **optimal PID controller design** are used. Controllers used for IPDT and FOIPDT can't be tuned by algorithm used for FOPDT.

Discussed Plant modeling will help in modeling of many industrial plant. And tuning method used for that plant will help to find out of controllers parameters. Response will suggest which tuning method is better for the plant. And also it will play great roll in selecting of controller.

# Reference

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- **Principal of measurement system – John P. Bentley**
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**Bequette**

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[.pdf](http://www.siam.org/books/dc14/DC14Sample.pdf)
- [http://en.wikipedia.org/wiki/PID\\_controller](http://en.wikipedia.org/wiki/PID_controller)

